

CMPE-242

Applied Feedback Control

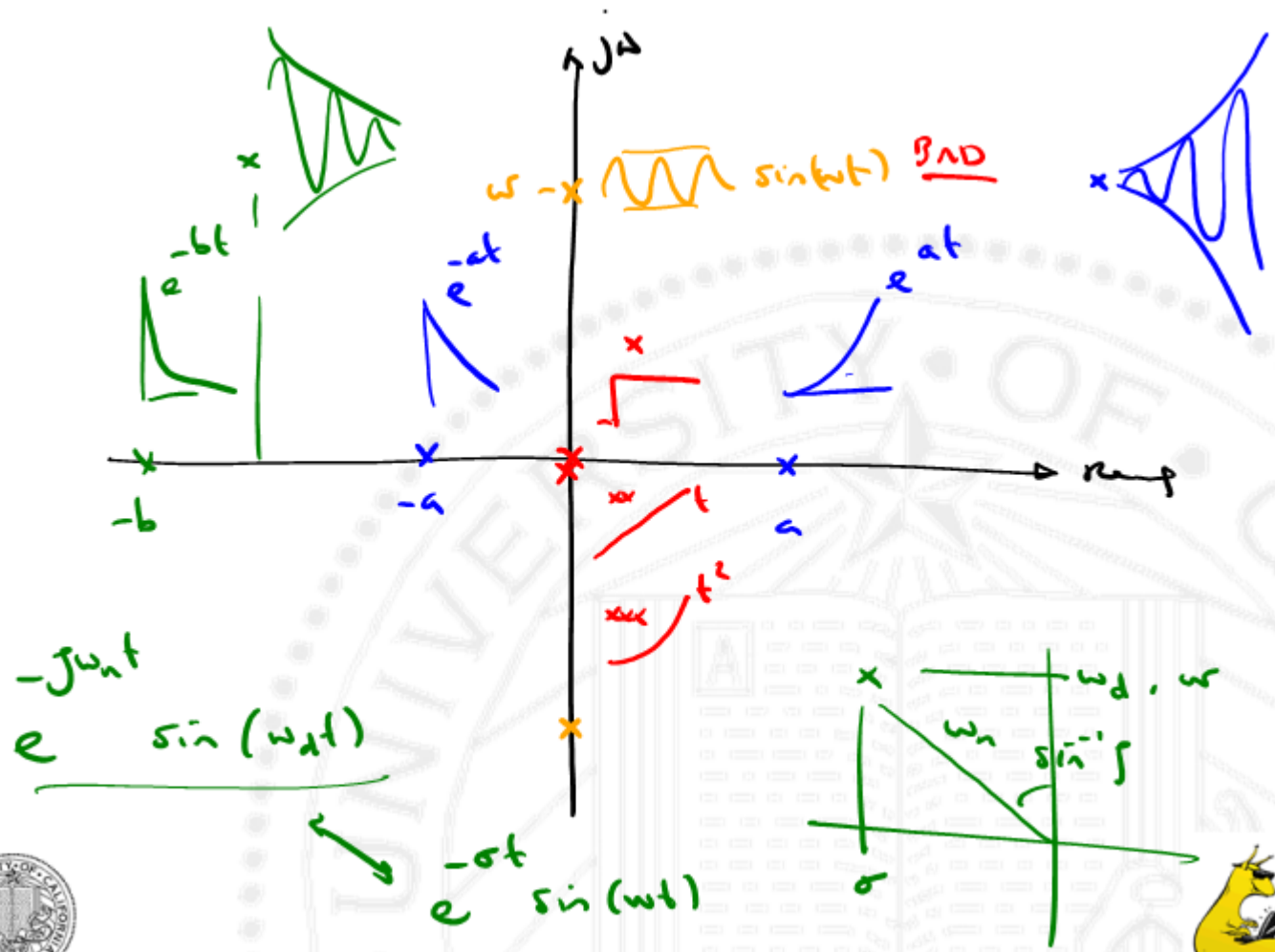
Gabriel Hugh Elkaim
Winter 2016

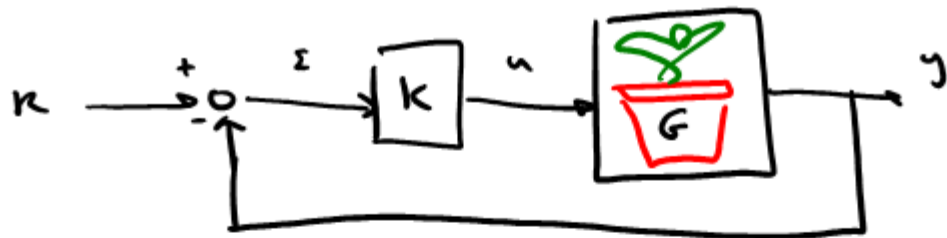


Announcements :

HW # 2 is posted.

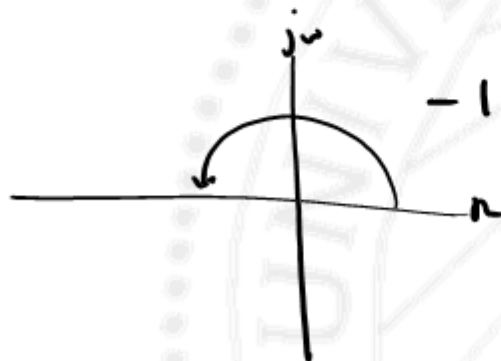






$$\frac{Y}{R} = \frac{GK}{[1 + GK]} \leftarrow \Delta(s) \text{ - characteristic equation}$$

Root locus: $\Delta(s) = 0 \rightarrow 1 + GK = 0 \therefore GK = -1$



$$-1 = 1e^{\pm \pi j}$$

-1 complex #: $|z| = 1$
 $\angle = \pm 180^\circ$

$$K(s) = K_0 \frac{N_L(s)}{D_L(s)}$$





$$\underline{F = G^{-1}}$$

$$K(s) = \frac{N_k}{D_k} \quad G(s) = \frac{N_g}{D_g} \quad \frac{Y}{R} = \frac{N_k N_g}{N_k N_g + D_g D_k} \leftarrow \Delta(s)$$

$$\frac{Y}{R} = \frac{N_g N_k + F N_k D_k}{N_k N_g + D_g D_k} \leftarrow \Delta(s)$$

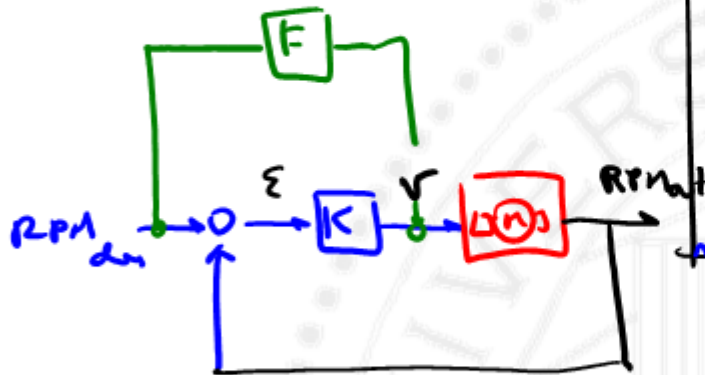


NO PLANT
INVERSION

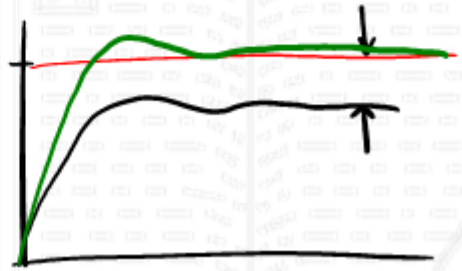
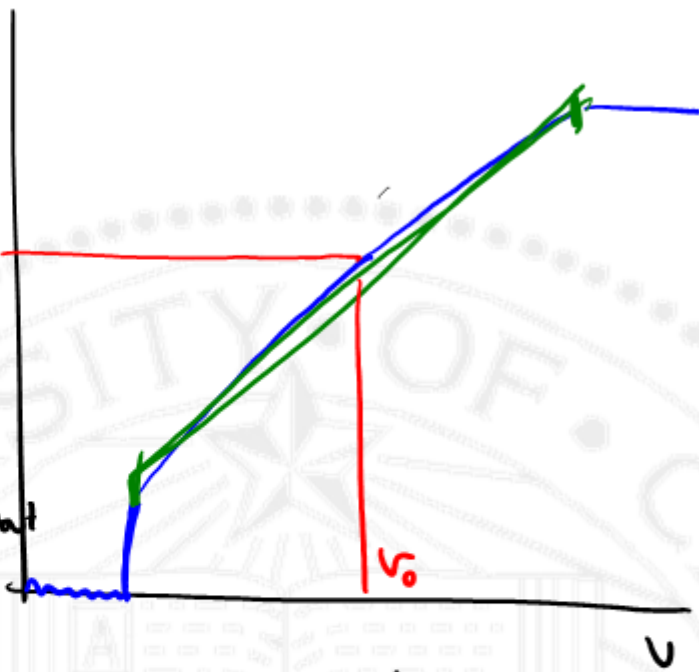


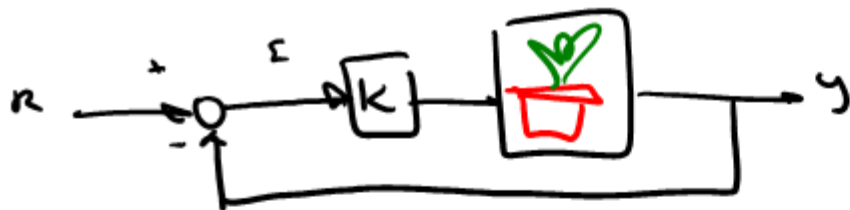
DC Motors

RPM



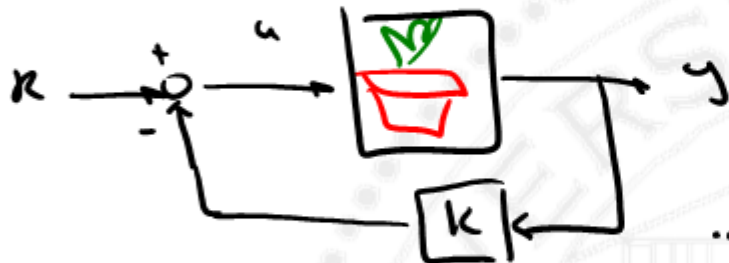
RPM_m





$$\frac{GK}{1+GK} \leftarrow \Delta(s)$$

"FM AUTOPILOT"



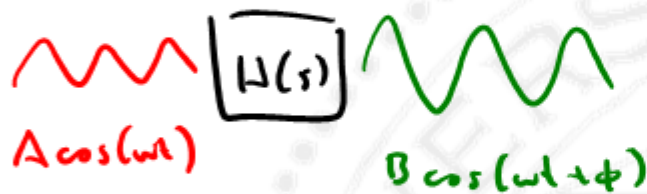
$$\frac{G}{1+GK} \leftarrow \Delta(s)$$

"STABILITY AUGMENTATION SYSTEM"



Root locus — move the poles of $\Delta(s)$ to get good "transient" response.

$H(s)$ — transfer function

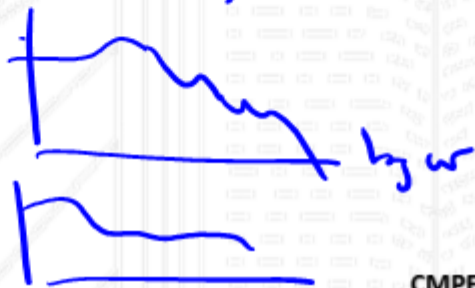


$H(j\omega)$ — complex number $\begin{matrix} | \cdot | \\ \angle \end{matrix}$

Bode

$\log \left| \frac{B}{A} \right|$

$\angle \phi$



$\frac{R}{C} \approx 1$ for some range of $j\omega$
"TRACKING RESPONSE"



STABILITY

BIBO — "Bounded Input / Bounded output"

$$|u(t)| \leq M < \infty$$



$$|y(t)| < N < \infty$$

$$H(s) = k \frac{\prod (\lambda - z_i)}{\prod (\lambda - p_i)}$$



All poles in the LHP

$$\boxed{\operatorname{Re}(p_i) < 0 \neq i}$$

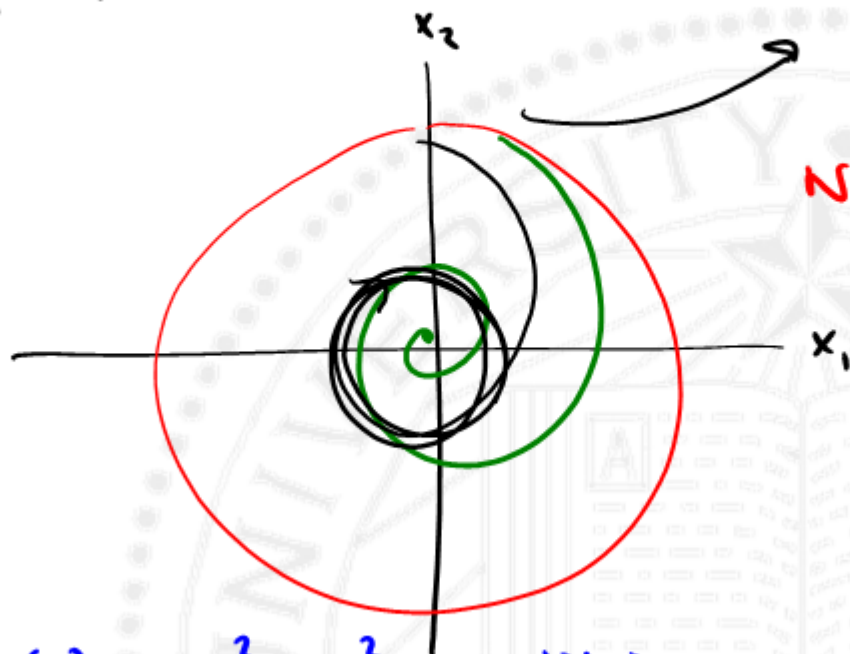
non-repeated root on $j\omega$ -axis



Lyapunov Stability

$$|x_{ic}| < M$$

$|x(t)| < N \forall t$ then stable



$N = \phi$ Asymptotically
stable

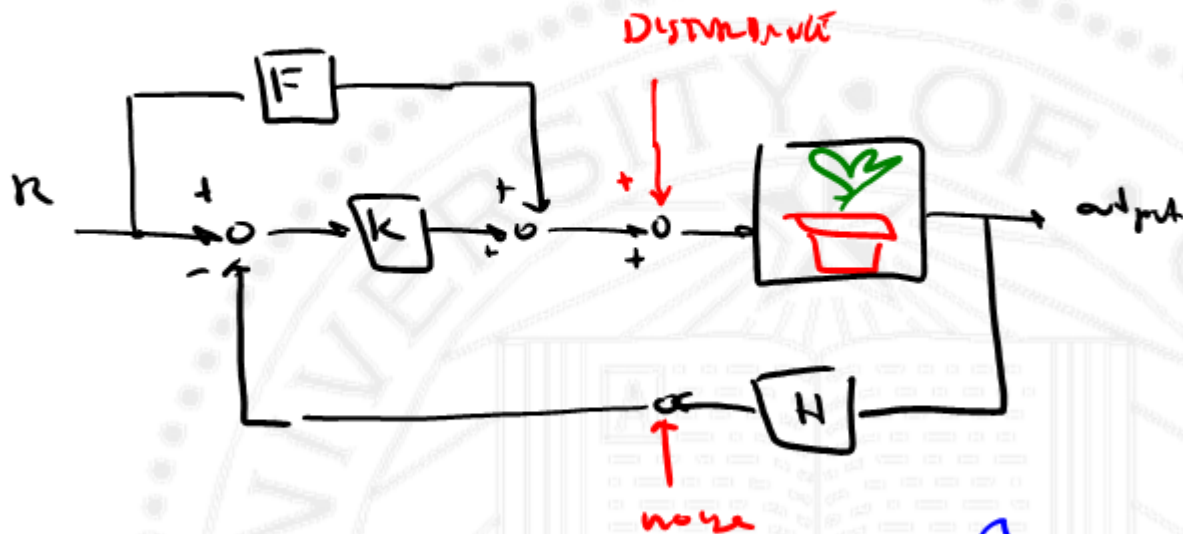
$$V(x) = x_1^2 + x_2^2$$

$$\frac{dV(x)}{dt} < \phi$$



Loop gains - TRANSIENT "KICK"

BODE - $|H(j\omega)| < \frac{1}{4}$ "TRACKING"



BE WARY OF LARGE GAINS



(1) STABILITY — all poles in LHP, Lyapunov, BIBO

(2) Overshoot
Rise Time
Steady State
Damping

TRANSIENT SPEC'S — ROOT LOCUS

(3) Bandwidth
Disturbance Rejection
Steady State Error

TRACKING RESPONSES — BODE

(4) Gain Margin
Phase Margin

ROBUSTNESS SPEC'S — NYQUIST

(5) Control Effort vs. Performance

cost trade-offs

$$\omega \propto \sqrt{y^2 + u^2}$$

"simulation"



x — position

$\frac{dx}{dt}$ — velocity

$\frac{d^2x}{dt^2}$ — acceleration

$\frac{d^3x}{dt^3}$ — jerk — "ride quality"

$\frac{d^4x}{dt^4}$ — snap — "comfort / fatigue"

x^5

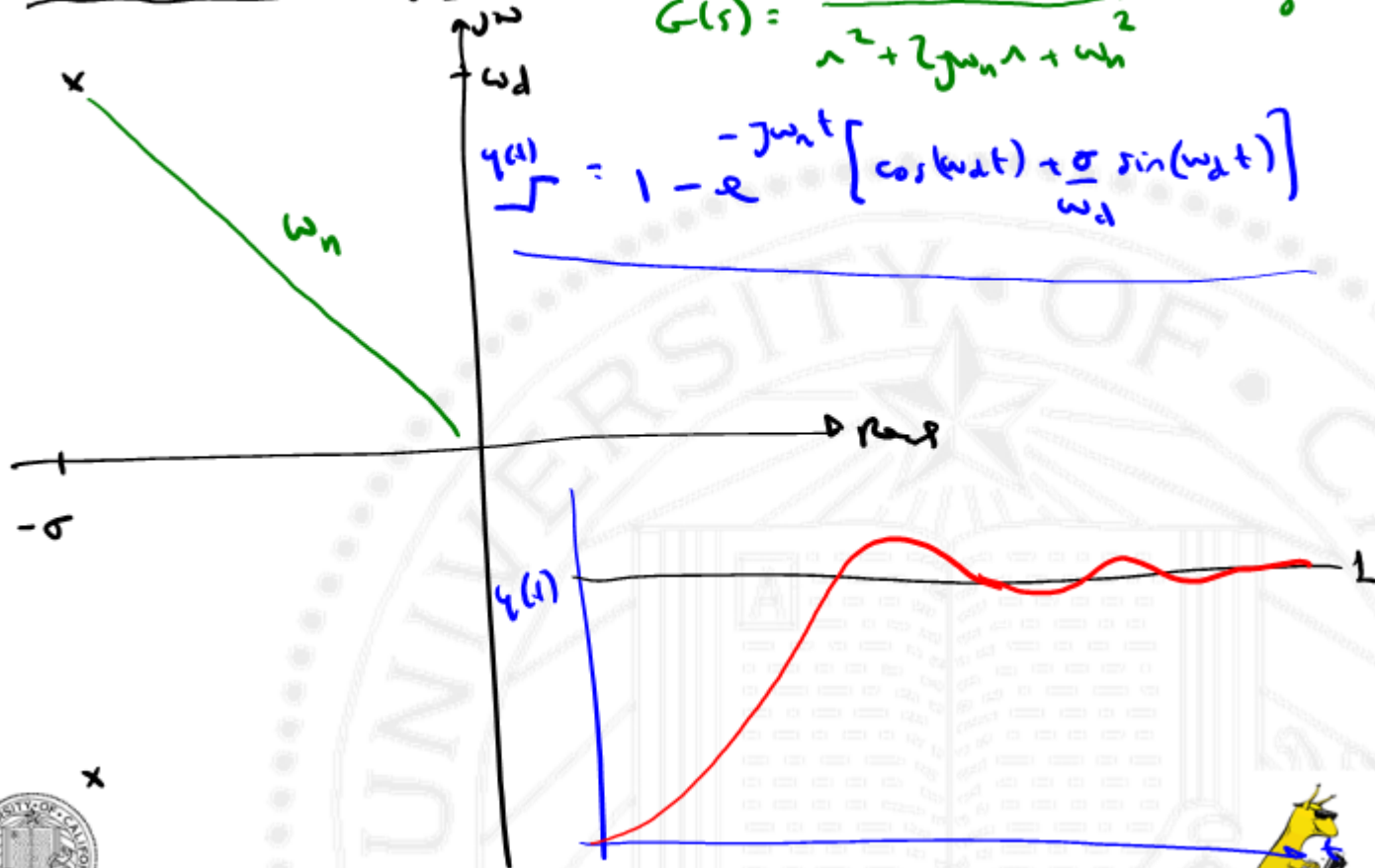


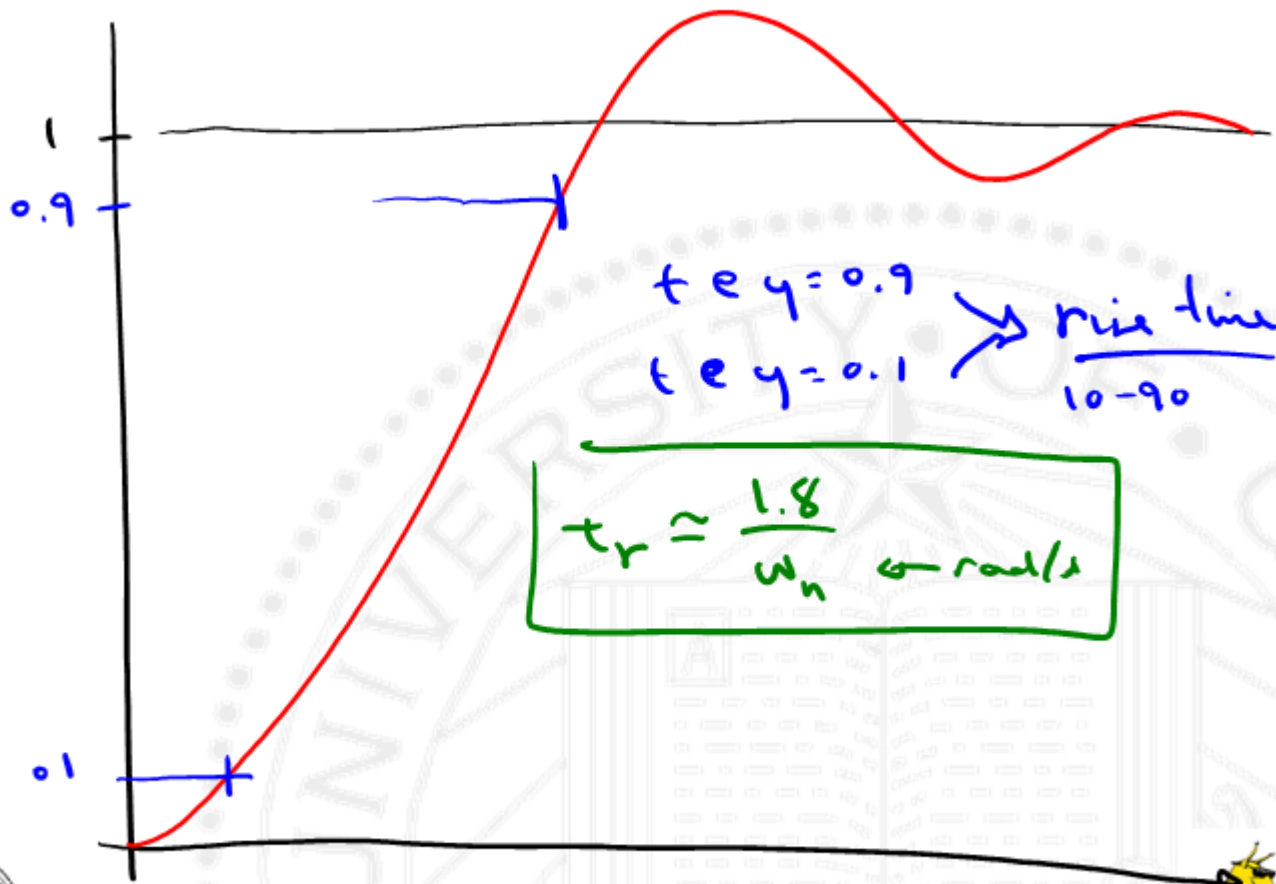
Performance Specs

$$G(s) = \frac{\omega_n^2}{s^2 + 2\zeta\omega_n s + \omega_n^2}$$

DC gain = 1

$$y(t) = 1 - e^{-\zeta\omega_n t} \left[\cos(\omega_d t) + \frac{\sigma}{\omega_d} \sin(\omega_d t) \right]$$

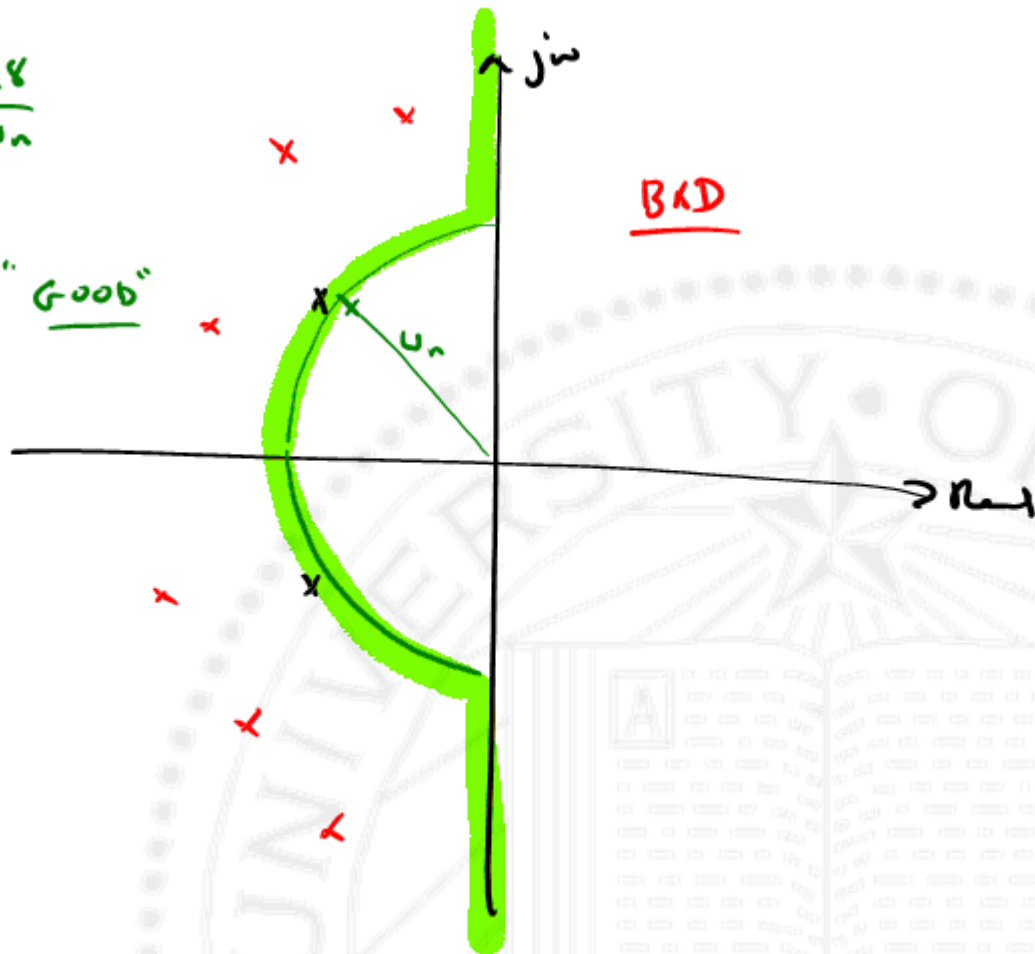


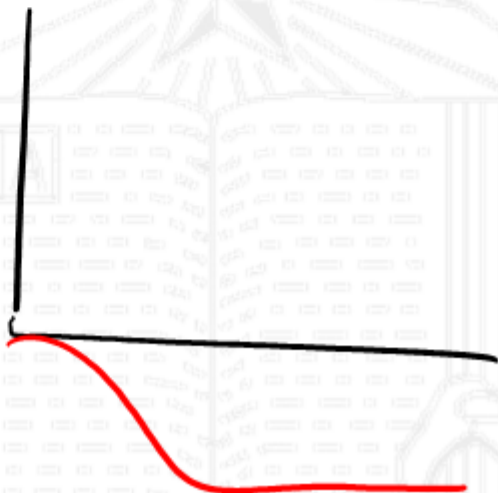
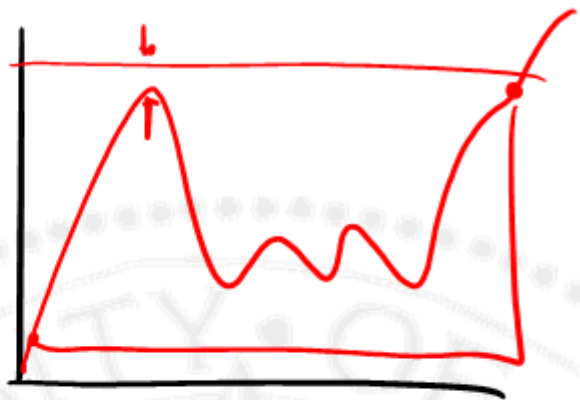
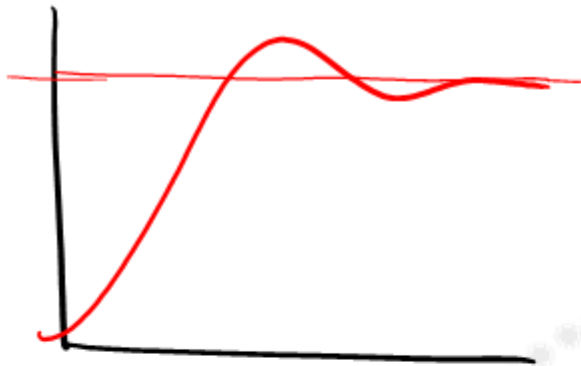


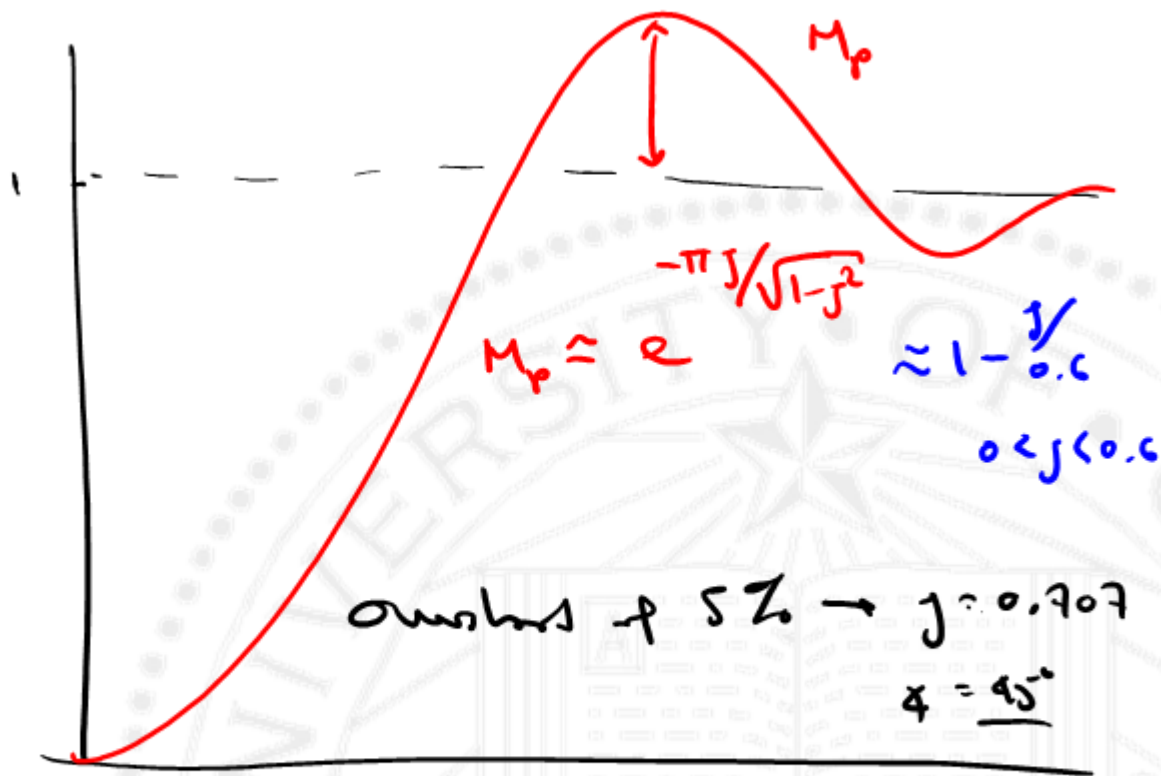
$$\tau_r \approx \frac{1.8}{\omega_n}$$

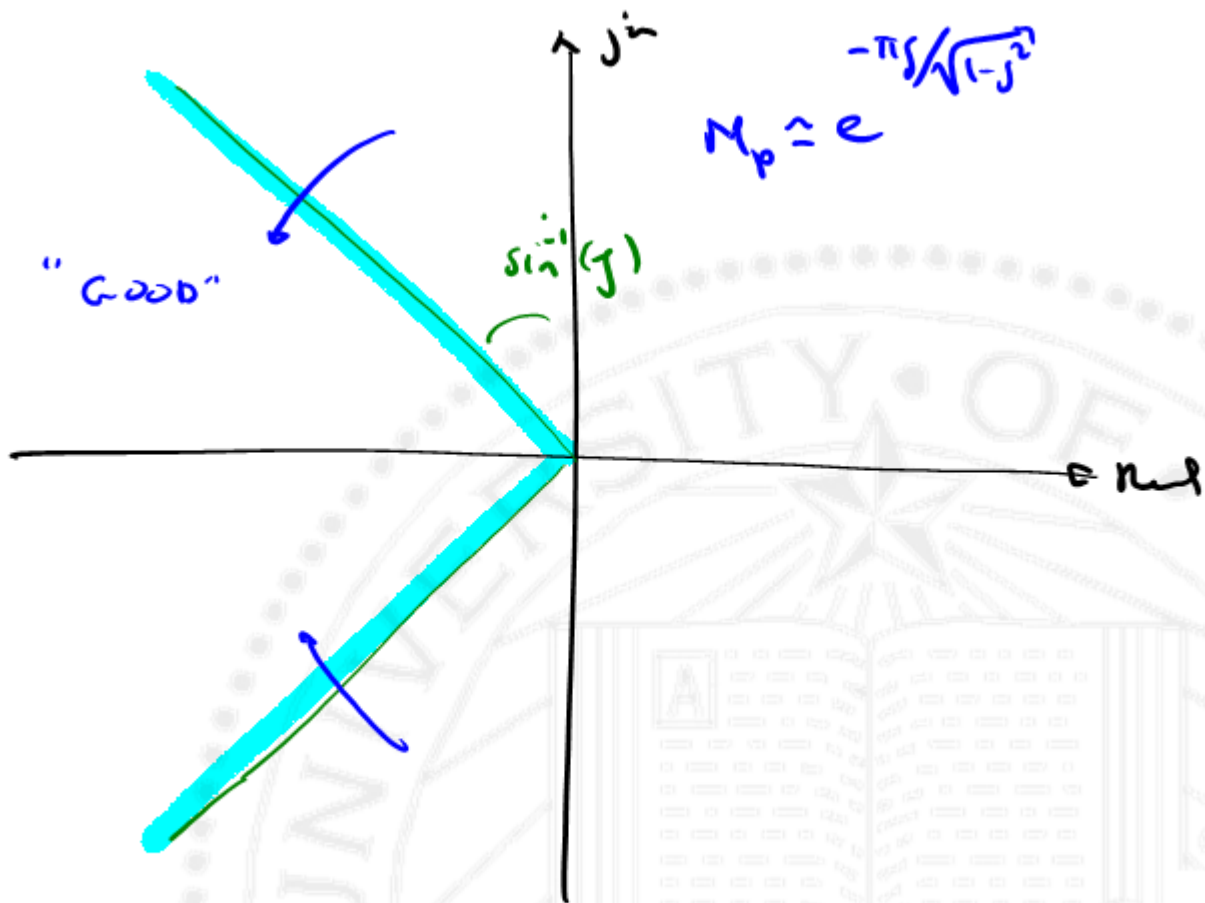
"Good"

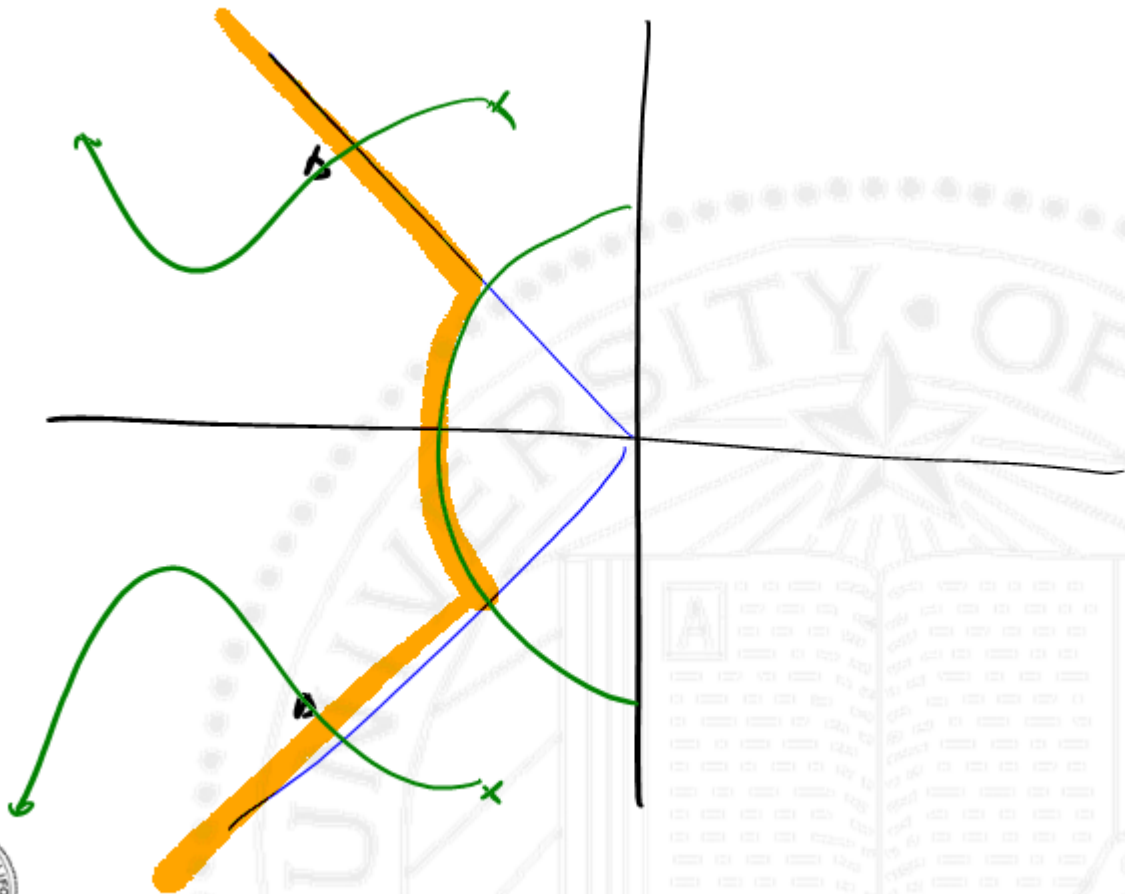
BAD

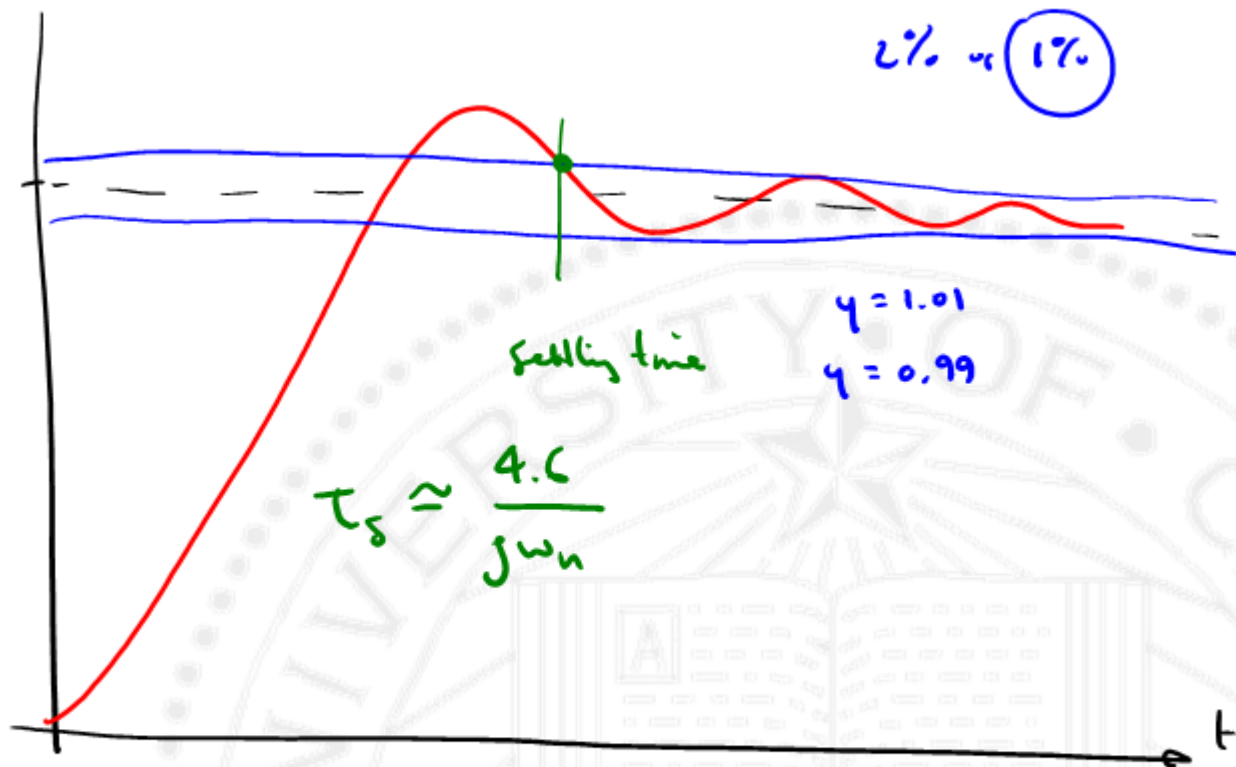








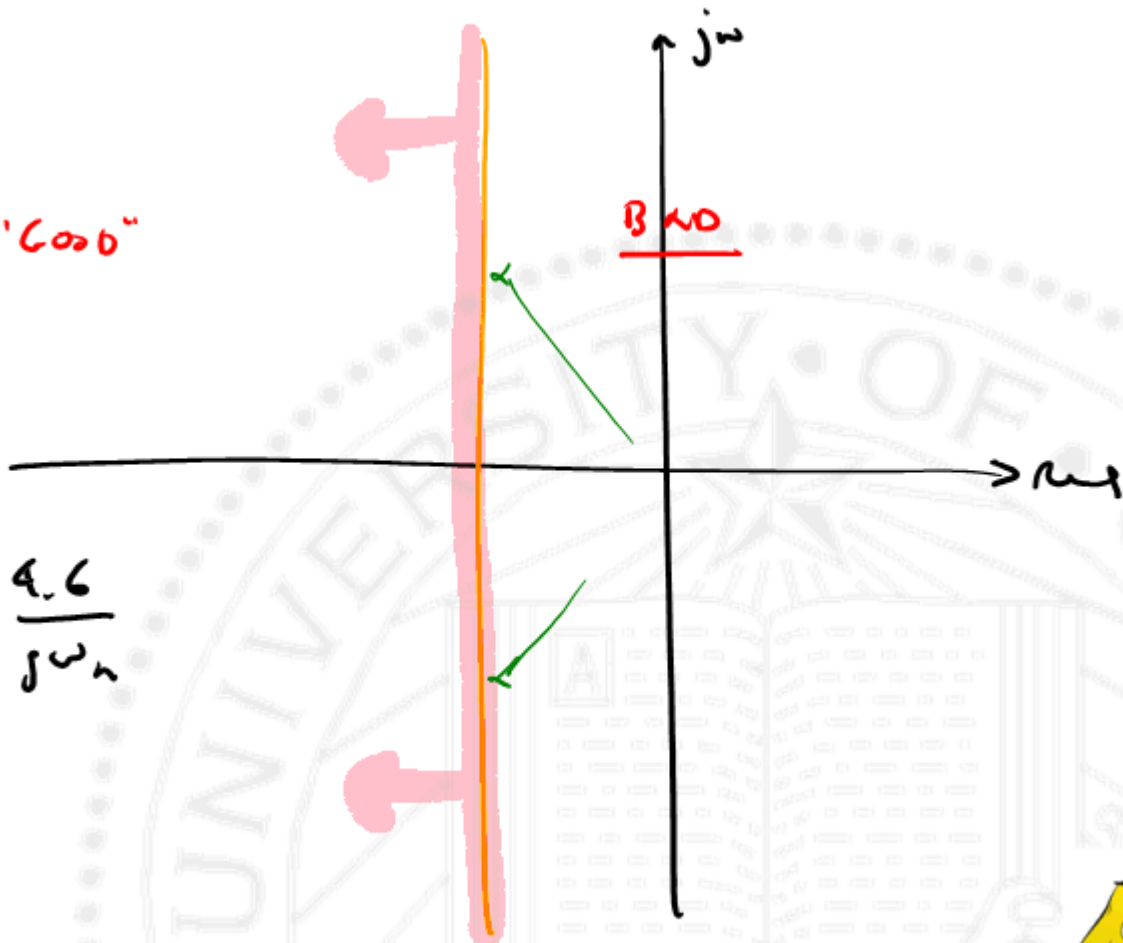




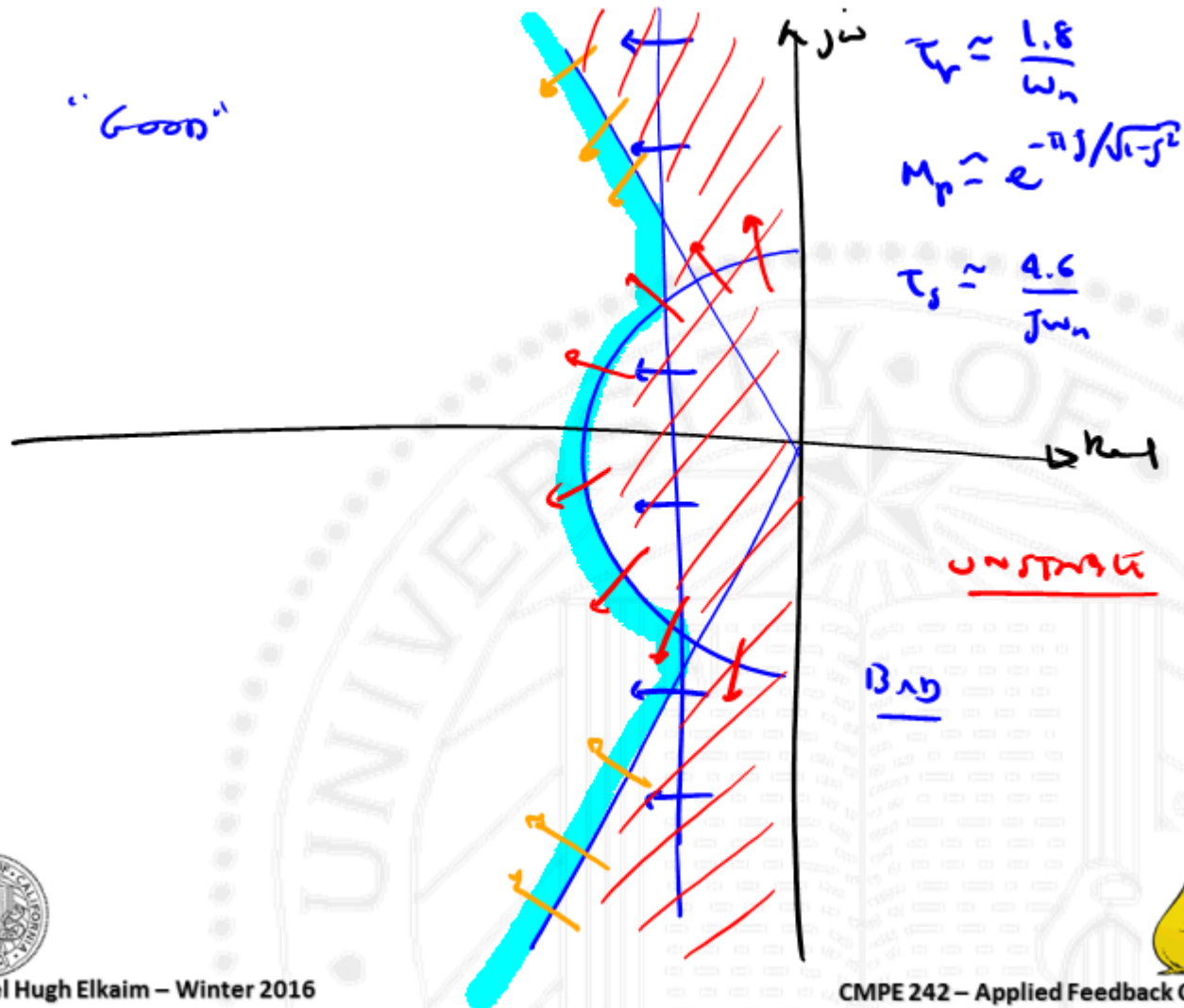
"Good"

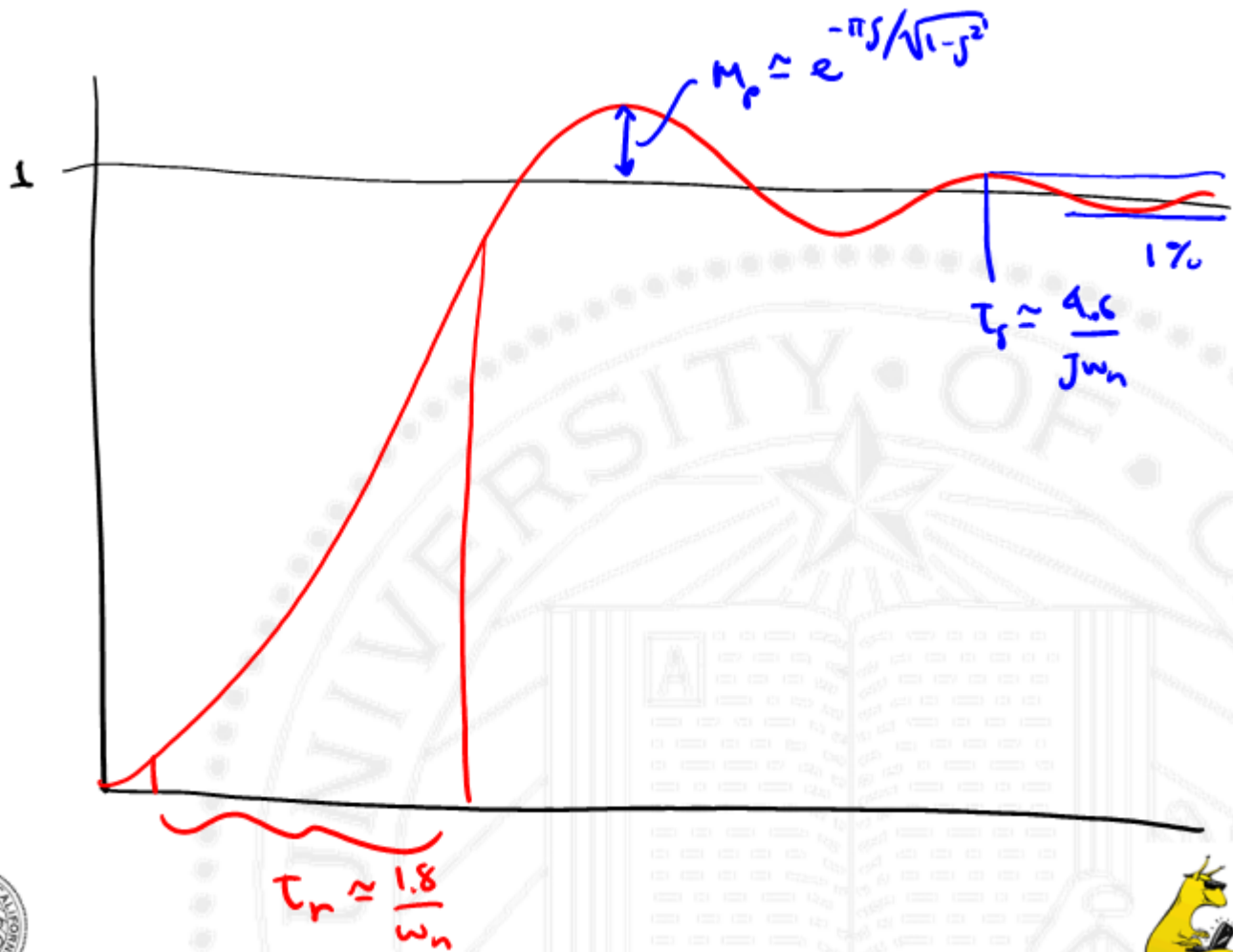
BND

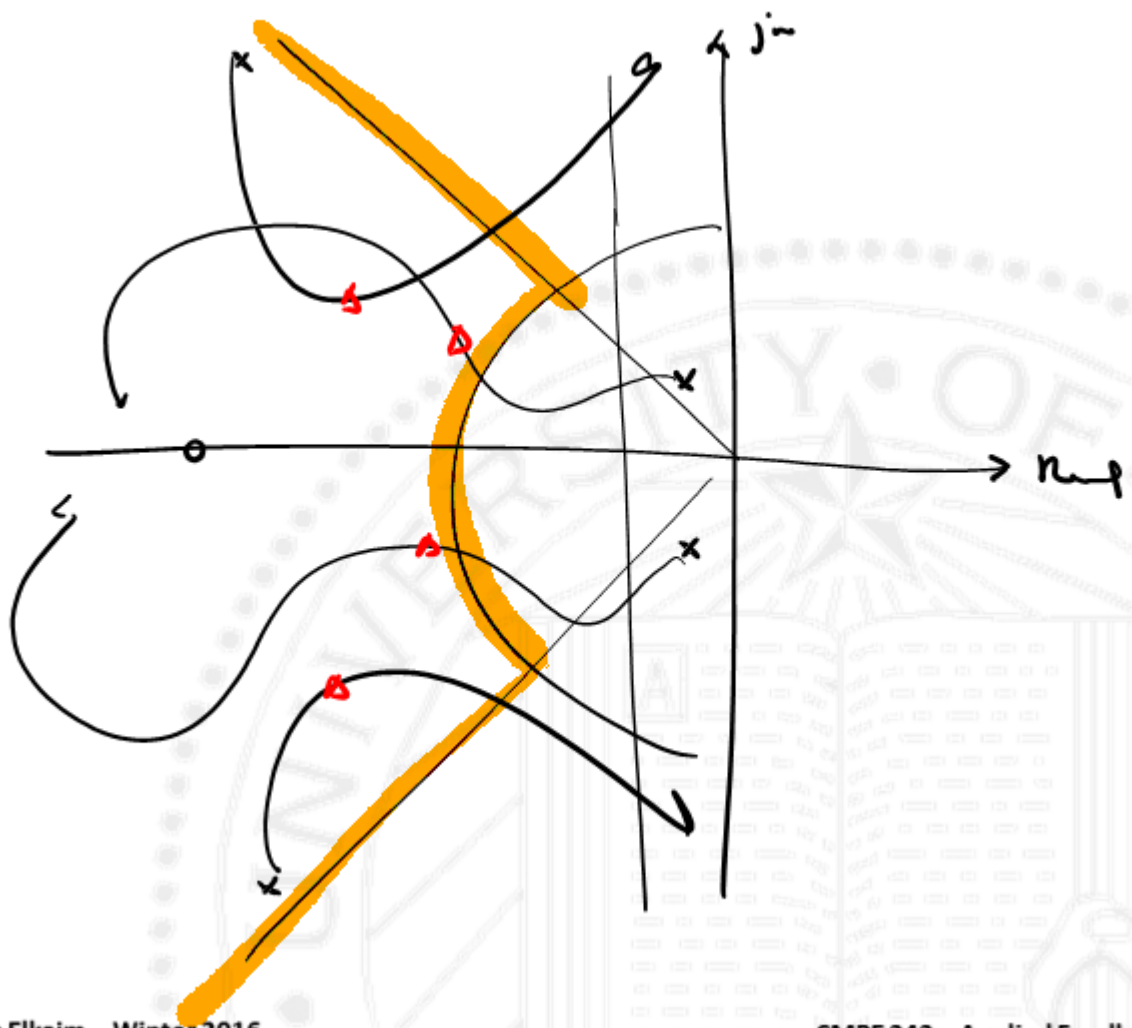
$$\tau_s \approx \frac{4.6}{\omega_n}$$



"Good"



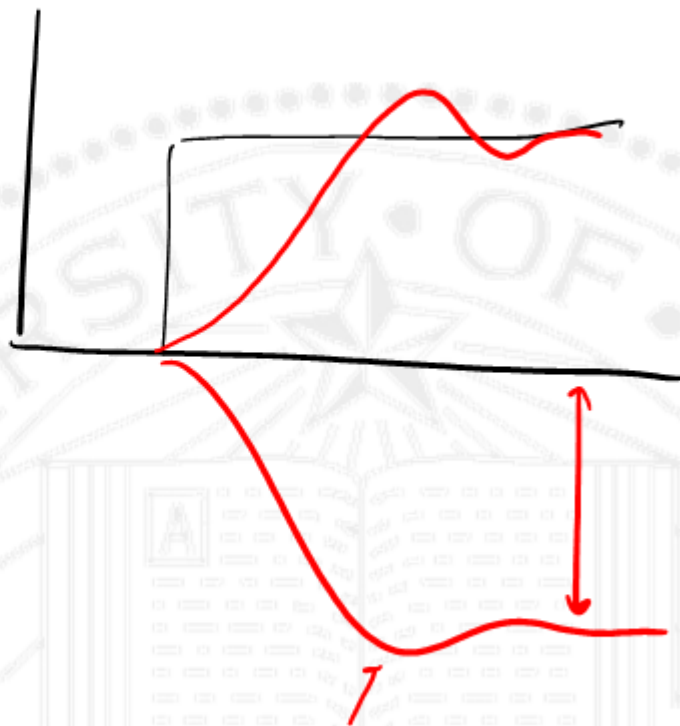


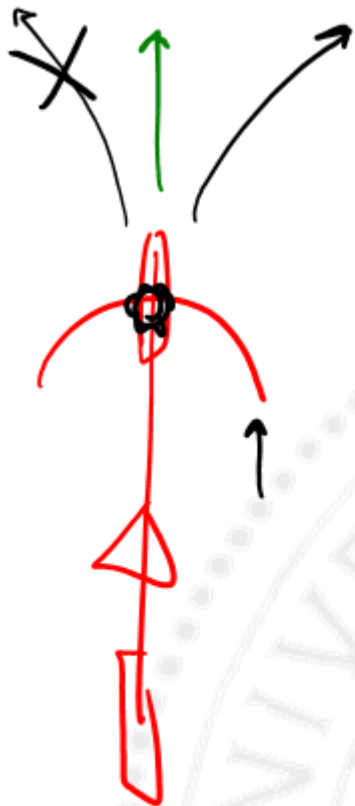


Poles of the system

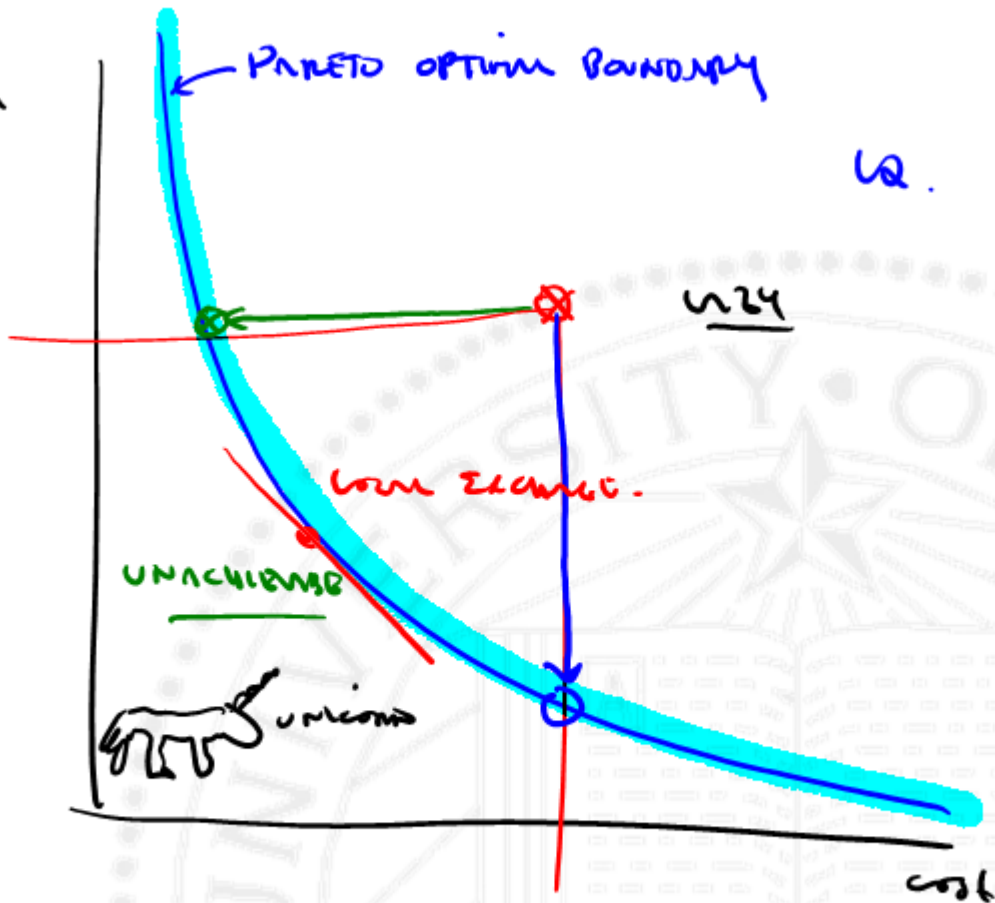
$$1 + GK = p \cdot - \Delta(s)$$

Zeros of the system





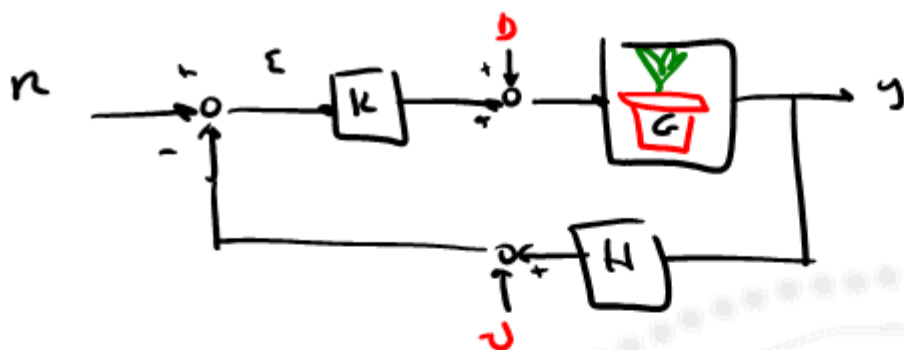
$\frac{1}{\text{performance}}$



u_{24}

u_{24}





$$H = 1$$

$$\frac{Y}{R} = \frac{GK}{1+GK} \quad \Delta(s) = \phi$$

$$\frac{\epsilon}{R} = \frac{1}{1+GK}$$

$$\frac{\epsilon_{ss}}{R} < \alpha \quad \text{"Thinking SBC"}$$

$$.1 > \frac{1}{1+GK} \quad \rightarrow \quad |1+GK| > 10 \quad |GK| > 11$$



$$\frac{Y}{R} = \frac{GK}{1+GK}$$

$GK \uparrow$ $|GK| BK \rightarrow$ TRACKING

$$\frac{E}{R} = \frac{1}{1+GK}$$

$GK \uparrow$ $|K| DK \rightarrow$ DISTURBANCE, REFLECTION

$$\frac{D}{X} = \frac{G}{1+GK}$$

$K \uparrow$

$$\frac{Y}{U} = \frac{GK}{1+GK}$$

$GK \downarrow$

GK small, good noise immunity



//

$\log(GK)$

