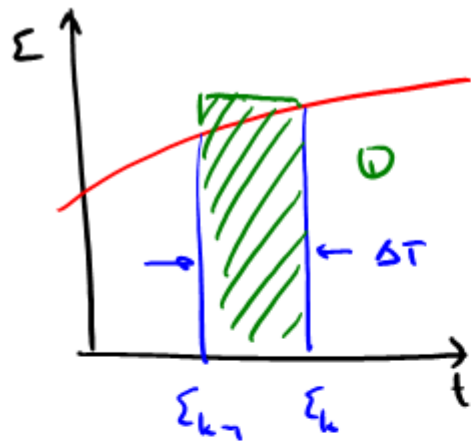


CMPE-242

Applied Feedback Control

Gabriel Hugh Elkaim
Winter 2016

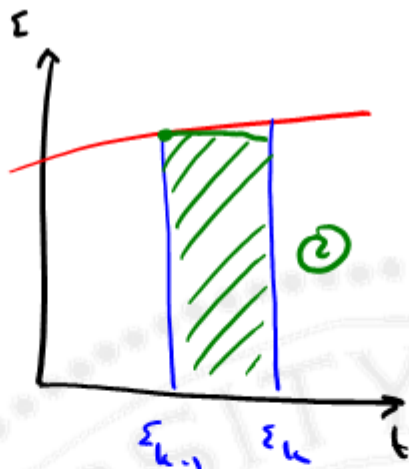




$$\Delta_k = \Delta_{k-1} + \epsilon_k \Delta T$$

$$\lambda = \frac{z^{-1}}{z \Delta T} \quad (1)$$

"BACKWARDS
EUREN"



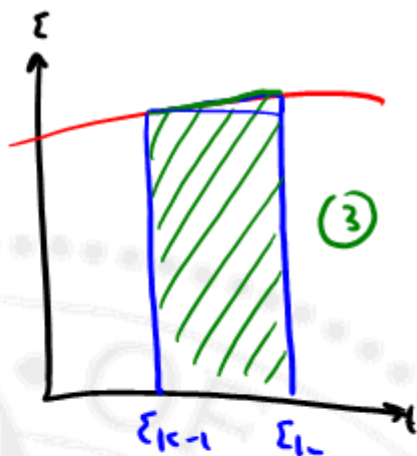
$$\Delta_k = \Delta_{k-1} + \epsilon_{k-1} \Delta T$$

$$(1-z^{-1})\Delta_k = z^{-1} \Delta T \epsilon_k$$

$$\frac{\Delta}{\epsilon} = \frac{z^{-1} \Delta T}{1-z^{-1}} = \frac{\Delta T}{z-1} = \frac{1}{\lambda}$$

$$\lambda = \frac{z-1}{\Delta T} \quad (2)$$

"Forward"



$$\Delta_k = \Delta_{k-1} + \Delta T \frac{(\epsilon_k + \epsilon_{k-1})}{2}$$

$$(1-z^{-1})\Delta_k = \frac{\Delta T}{2} (1+z^{-1}) \epsilon_k$$

$$\frac{\Delta}{\epsilon} = \frac{\Delta T}{2} \frac{(1+z^{-1})}{(1-z^{-1})} = \frac{1}{\lambda}$$

$$\lambda = \frac{2(z-1)}{\Delta T (z+1)} \quad (3)$$

"TUDIN"



$$\textcircled{1} \frac{1}{\lambda} = z \frac{\Delta T}{z^{-1}} \leftrightarrow \lambda = \frac{z^{-1}}{z \Delta T} \therefore z = \frac{1}{1 - \Delta T \lambda} \quad \text{"Backwards"}$$

$$\textcircled{2} \frac{1}{\lambda} = \frac{\Delta T}{z^{-1}} \leftrightarrow \lambda = \frac{z^{-1}}{\Delta T} \therefore z = 1 + \Delta T \lambda \quad \text{"Forward"}$$

$$\textcircled{3} \frac{1}{\lambda} = \frac{\Delta T}{2} \frac{(z+1)}{(z-1)} \leftrightarrow \lambda = \frac{2}{\Delta T} \frac{(z-1)}{(z+1)} \therefore z = \frac{1 + \frac{\Delta T}{2} \lambda}{1 - \frac{\Delta T}{2} \lambda} \quad \text{"Tustin"}$$

$$\textcircled{4} e^{-\Delta T \lambda} = z^{-1} \leftrightarrow \lambda = \frac{1}{\Delta T} \ln(z) \therefore z = e^{\Delta T \lambda} \quad \text{"Exact"}$$



$$\textcircled{4} \quad z = e^{\Delta T \lambda} = \underline{1} + \underline{\Delta T \lambda} + \underline{\frac{\Delta T^2}{2!} \lambda^2} + \underline{\frac{\Delta T^3}{3!} \lambda^3} + \dots$$

$$\textcircled{3} \quad z = \frac{1 + \frac{\Delta T}{2} \lambda}{1 - \frac{\Delta T}{2} \lambda} = \underline{1} + \underline{\Delta T \lambda} + \underline{\frac{\Delta T^2}{2} \lambda^2} + \underline{\frac{\Delta T^3}{4} \lambda^3} + \dots$$

$$\textcircled{2} \quad z = 1 + \Delta T \lambda = \underline{1} + \underline{\Delta T \lambda} + \phi$$

$$\textcircled{1} \quad z = \frac{1}{1 - \Delta T \lambda} = \underline{1} + \underline{\Delta T \lambda} + \underline{\Delta T^2 \lambda^2} + \underline{\Delta T^3 \lambda^3} + \dots$$



$$K(s) = K \frac{s+a}{s+b} \Big|_{s = \frac{1}{sT} \ln(z)} = \frac{K \left(\frac{1}{sT} \ln(z) + a \right)}{\frac{1}{sT} \ln(z) + b}$$

"hard to implement"

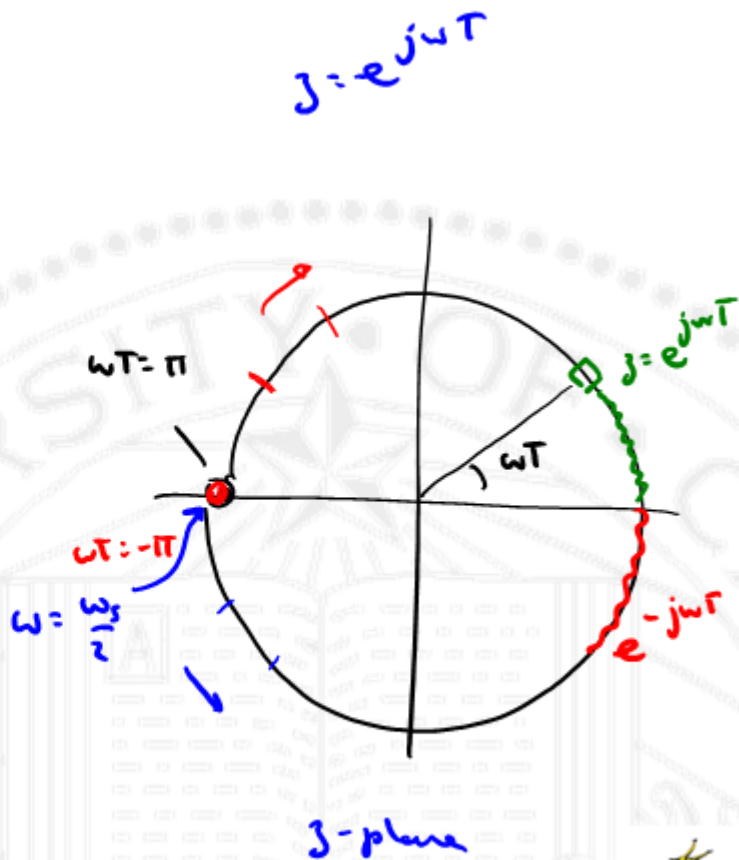
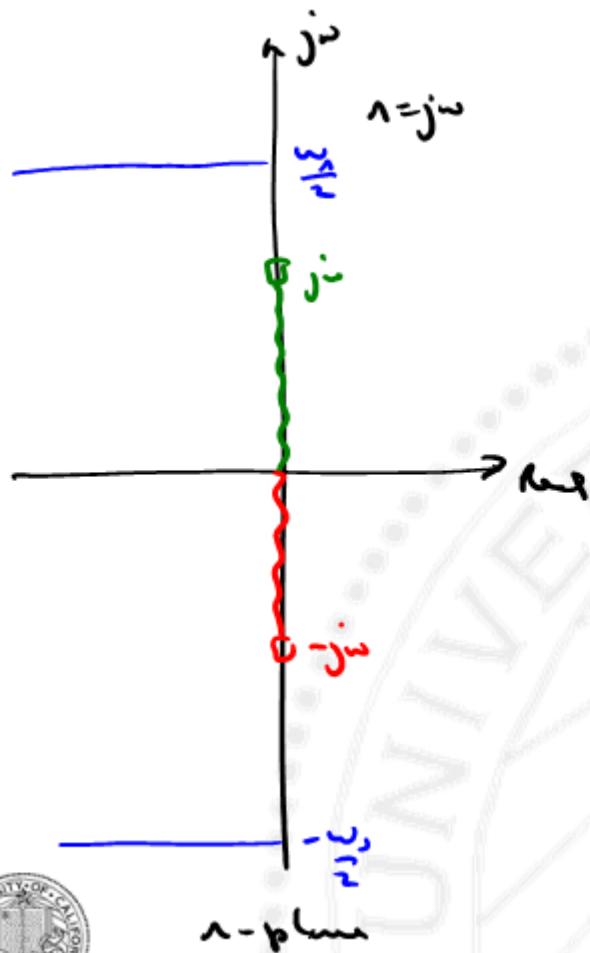
Exact is useful \rightarrow equivalent poles are?

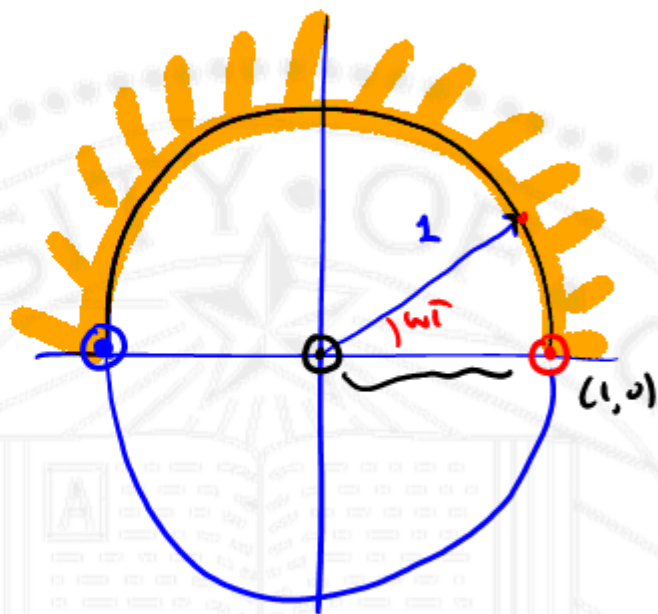
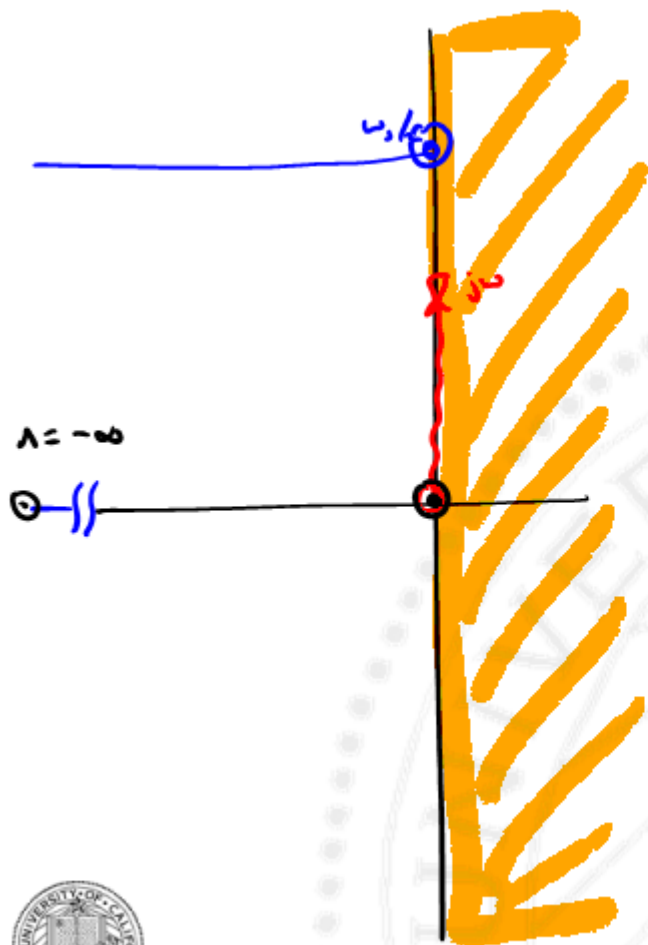
ODE \rightarrow $K(s)$ $s = j\omega$ — Bode Plot $\ll \ll$

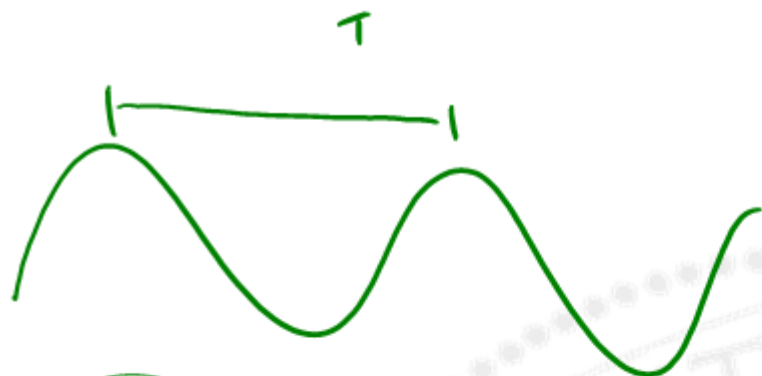
ODE \rightarrow $K(z)$ $z = e^{sT}$ — pole/zeros

$K(z)$ $\Big|_{z = e^{j\omega T}}$ — \ll, \ll (dBode)





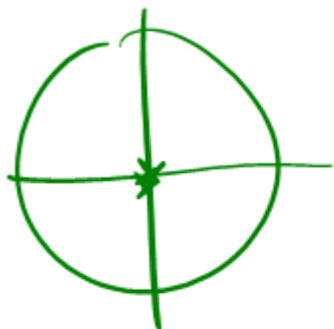




$$\omega_s = \frac{2\pi}{T}$$

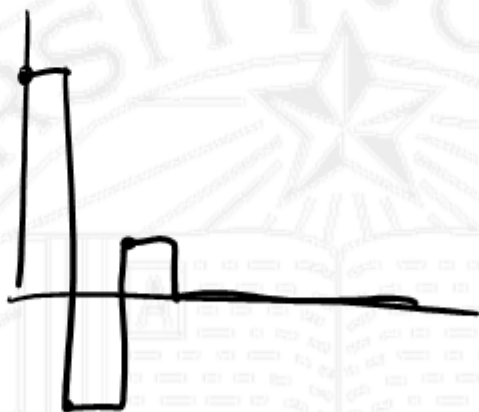
$\omega_s = \frac{3.14}{T}$ Min Freq.

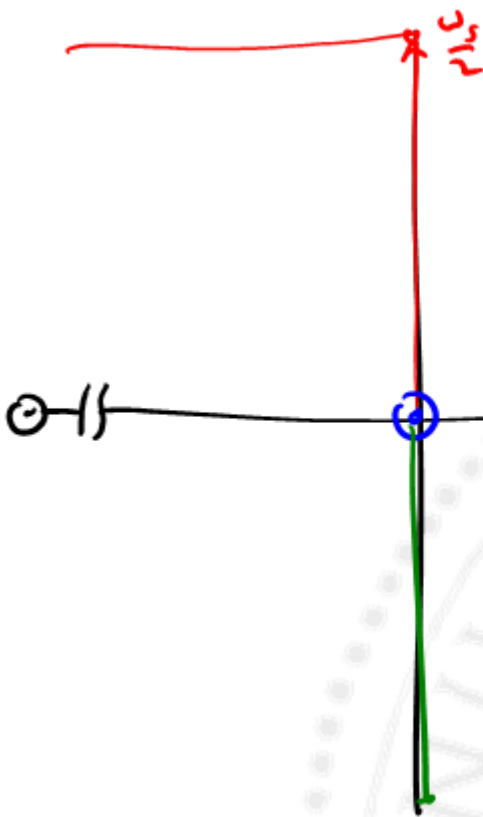




"Deadbeat Control"

n^{th} order plant will go to zero in n steps





$$\frac{\Sigma}{u} = z + \frac{1}{2} \rightarrow \Sigma_k = 3u_k + \frac{1}{2}u_k$$

$$\Sigma_k = u_{k+1} + \frac{1}{2}u_k$$

$$u_{k+1} = \Sigma_k - \frac{1}{2}u_k$$

$$\Sigma_k = 0$$

$$\Sigma_0 = 1 \quad u_0 = 0$$

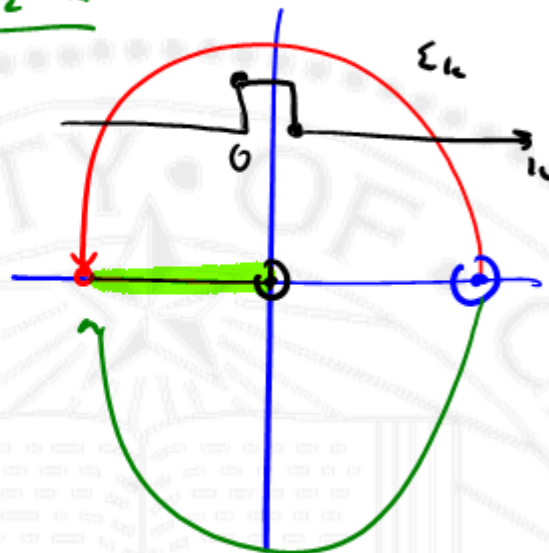
$$\Sigma_1 = 0 \quad u_1 = 1$$

$$u_2 = -\frac{1}{2}$$

$$u_3 = \frac{1}{4}$$

$$u_4 = -\frac{1}{8}$$

...



$$n = \frac{1}{T} \ln\left(\frac{1}{2}\right) \quad z = -\frac{1}{2} \quad z = \frac{1}{2} e^{\pm \pi j}$$



$$n = \frac{1}{T} \ln\left(\frac{1}{2} e^{\pm \pi j}\right) = \frac{1}{T} \ln\left(\frac{1}{2}\right) \ln\left(e^{\pm \pi j}\right)$$

$$= \frac{1}{T} \ln\left(\frac{1}{2}\right) \pm \pi j$$

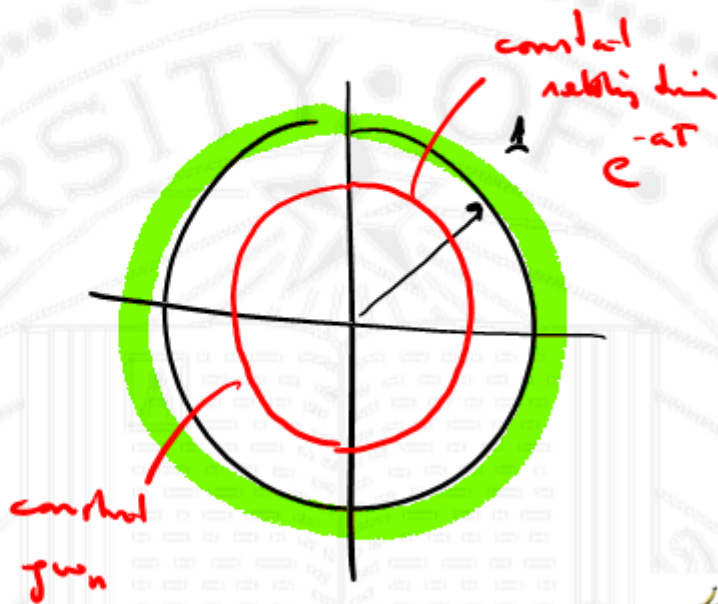


কর কসর



$$\lambda = \frac{1}{T} \ln(z) \iff z = e^{T\lambda}$$

$$\lambda = -a + bj \iff z = e^{(-a + bj)T} = e^{-aT} e^{jbT}$$



STABILITY \iff INSIDE UNIT CIRCLE



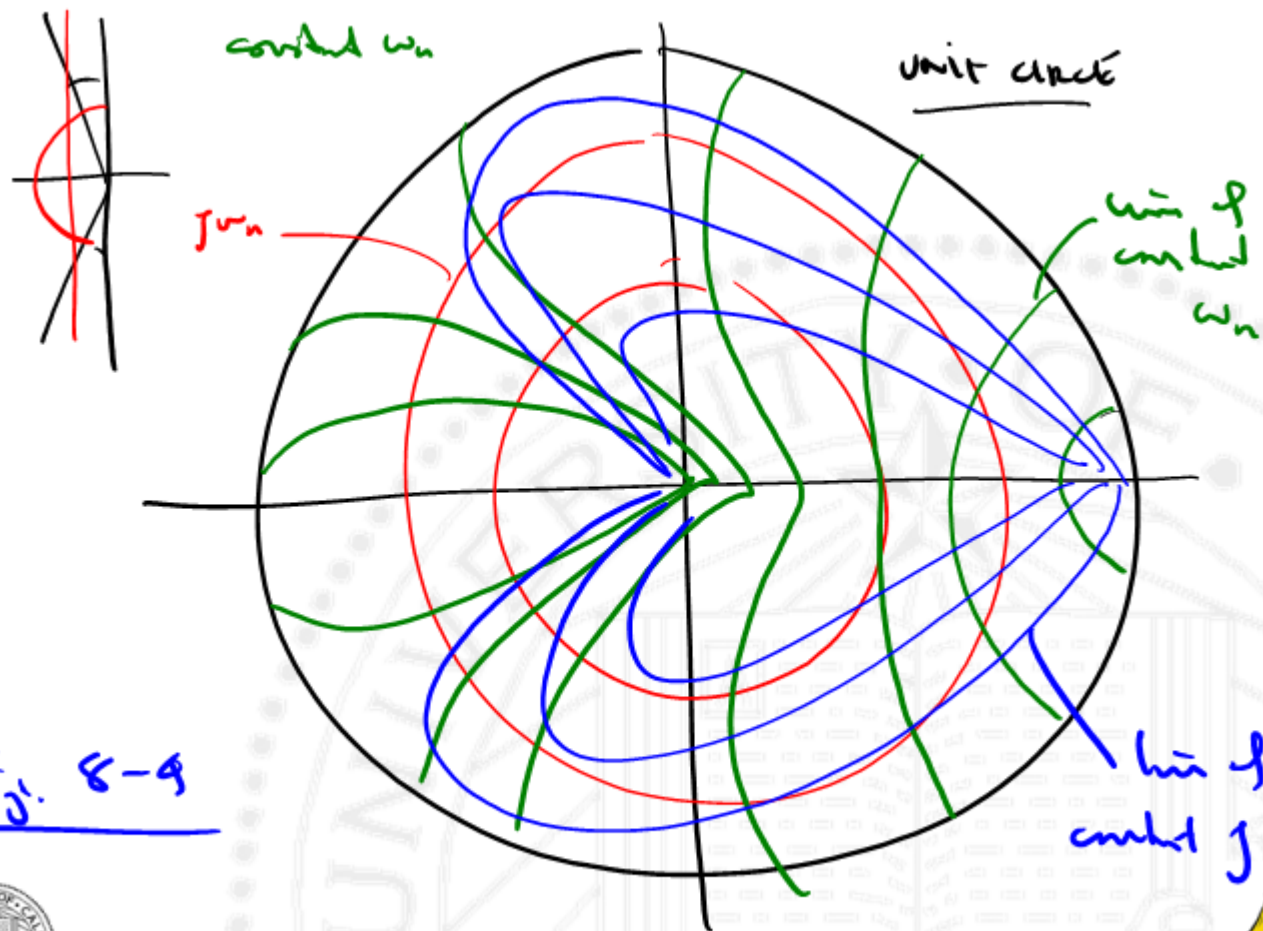
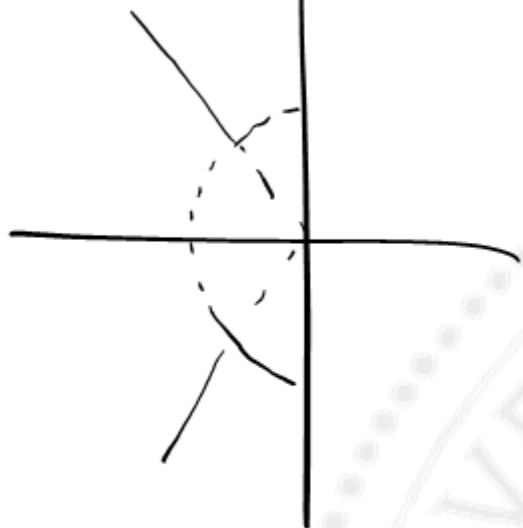


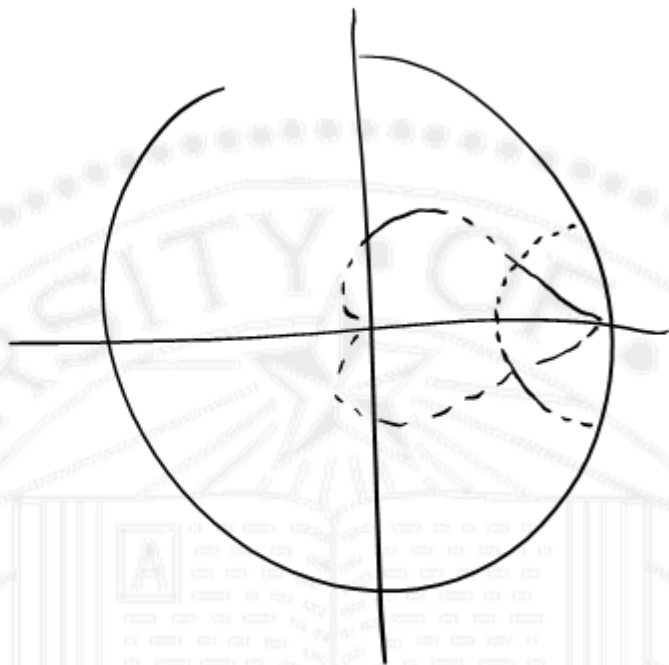
Fig. 8-9

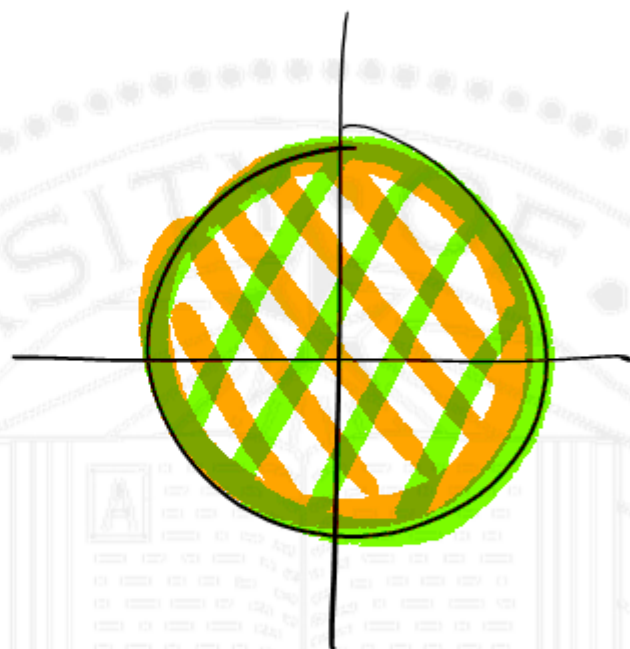
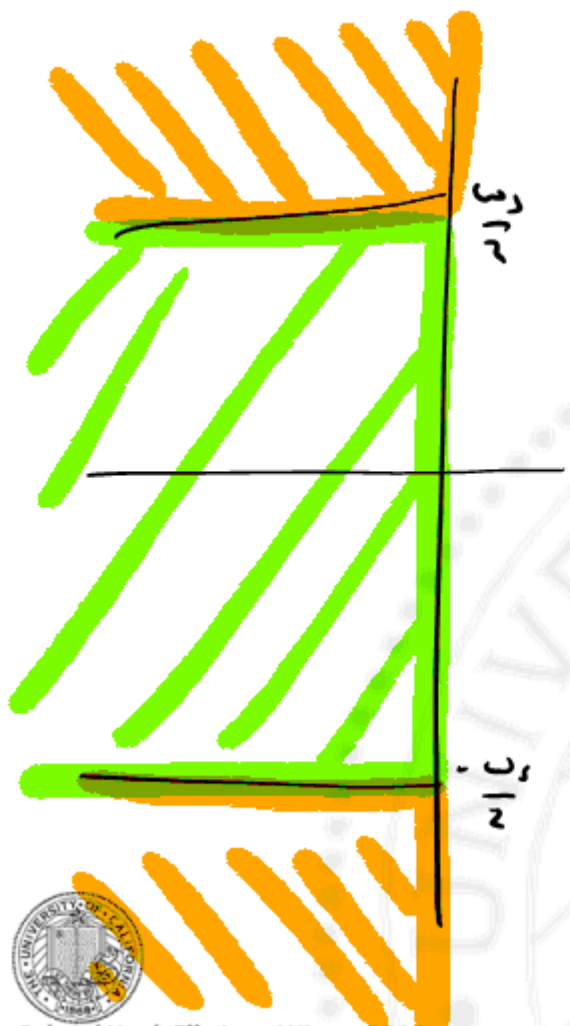


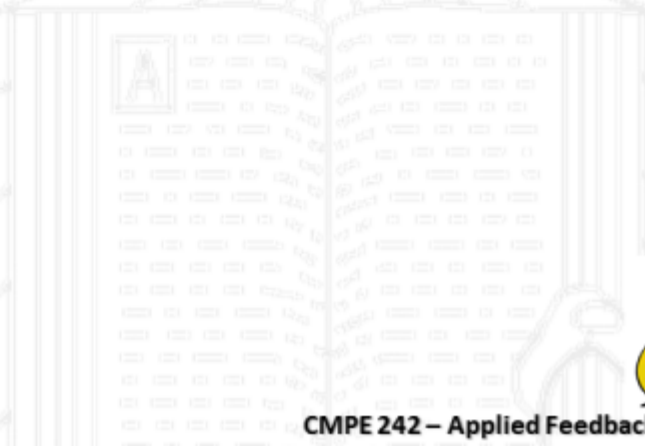
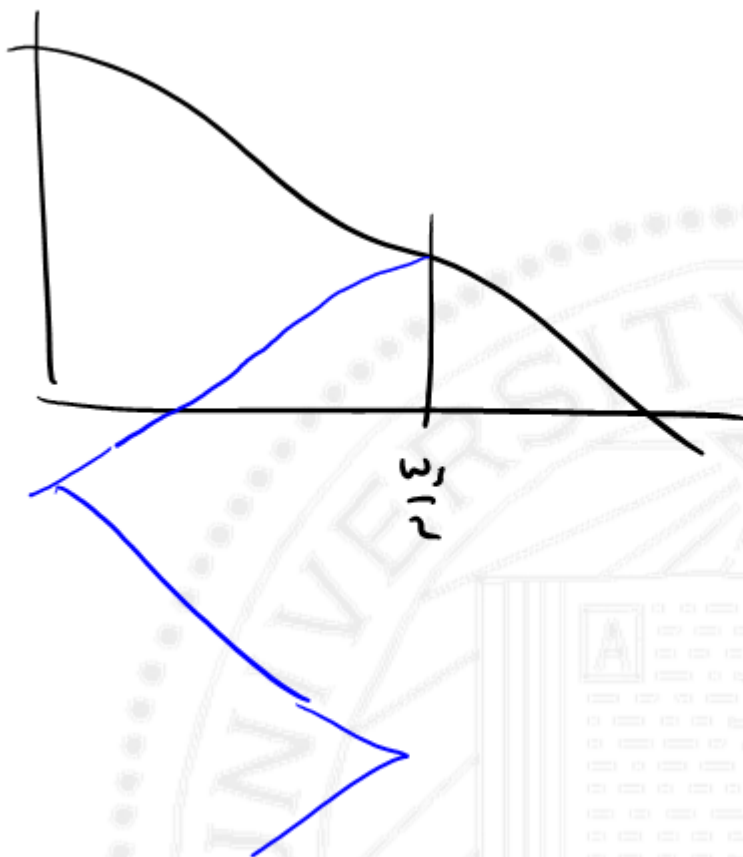
Sgrid (ω_n, J)

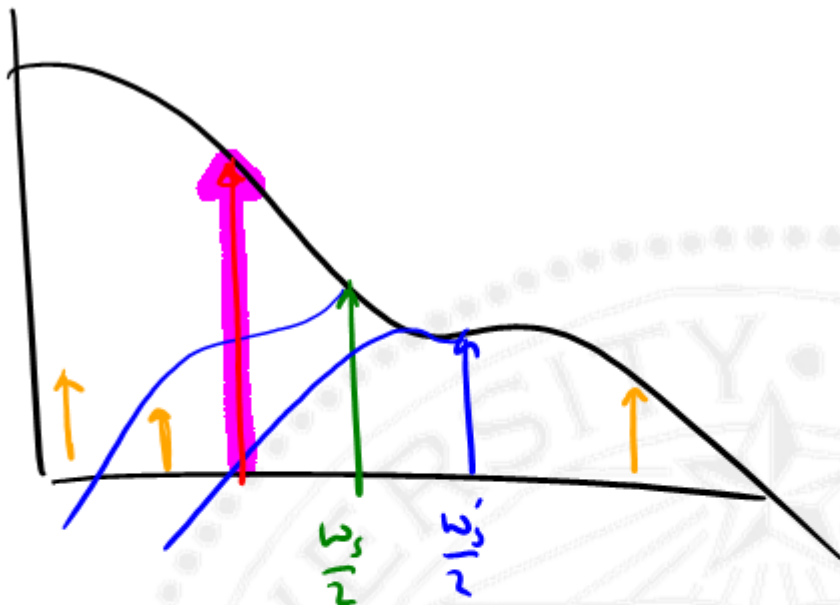


zgrid (ω_n, J)









not multiples of same other Final.



Design in analog world \sim account for $\frac{\sigma_T}{2}$ delay

convert $K(s) \rightarrow K(z)$ | JUSTIN

$$r = \frac{1}{\sigma_T} \ln(z)$$



$$K(s) = \frac{a}{s+ca}$$



$$K(j\omega) = \left. \frac{a}{s+ca} \right|_{s=j\omega} = \frac{a}{\frac{j\omega}{T} + a} = \frac{aT}{j\omega - 1 + aT} = \frac{aT}{j\omega - (1-aT)} \quad \text{"Forward's"}$$

$$K(j\omega) = \left. \frac{a}{s+ca} \right|_{s=j\omega} = \frac{a}{\frac{j\omega}{T} + a} = \left(\frac{aT}{1+aT} \right) \frac{1}{j\omega - \left(\frac{1}{1+aT} \right)} \quad \text{"Poles/zeros"}$$

$$K(j\omega) = \left. \frac{a}{s+ca} \right|_{s=j\omega} = \frac{a}{\frac{j\omega}{T} + a} = \left[\frac{aT}{1+aT} \right] \frac{1}{j\omega - \left(\frac{1-aT/2}{1+aT/2} \right)}$$



DIGITAL CONTROL VIA MAPPING

(1) Account for $\frac{\Delta T}{2}$ delay $\left\{ \begin{array}{l} \text{Bode} \sim \Delta\phi = -\frac{\omega T}{2} \\ \text{RL} \sim \text{Phase} (0^\circ) \end{array} \right.$

(2) Sample @ 20-30x ω_{x0} (ω_n)

(3) Design $K(s)$ w/ analog tools

(4) $K(z) \Leftarrow K(s)$ w/ TUSTIN

(5) Specific freq of interest (ω_{x0}) \rightarrow Prewarp



Prewarping

$|K(s)|$ — "Notch"

↓ Tustin

$K(z)$

$\omega_1 \neq \omega_2$

ω_1

$|K(s)|$
 $s = j\omega$

$K(z)$
 $z = e^{j\omega T}$

ω_2



$$K(s) \rightarrow K(\bar{s})$$

$$z = j\omega_a T = \frac{z}{T} \frac{e^{j\omega_1 T} - 1}{e^{j\omega_1 T} + 1}$$

$$e^{j\omega_1 T} = \cos(\omega_1 T) + j\sin(\omega_1 T)$$

$$j\omega_a T = \frac{z}{T} \tan\left(\frac{\omega_1 T}{2}\right)$$

$$K(z) = K(s) \Big|_{s = \left[\frac{\omega_1}{\tan\left(\frac{\omega_1 T}{2}\right)} \right] \frac{z-1}{z+1}}$$

prewarping

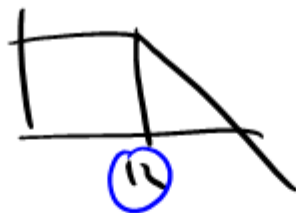


Tustin w/ Prewarping

$$s = \left[\frac{\omega_s}{\tan\left(\frac{\omega_s T}{2}\right)} \right] \frac{z-1}{z+1}$$



$$K(s) = \frac{12}{s+12}$$



$$28 \text{ Hz} = T = \frac{1}{25}$$

$$K(\omega) = K(s) \Big|_{s = \left(\frac{\omega}{\tan(\frac{\omega T}{2})} \right)^{\frac{-1}{s+1}}} = \frac{12}{s+12} \Big|_{s = \left(\frac{12}{\tan(\frac{12/25}{2})} \right)^{\frac{-1}{s+1}}}$$

$$K(\omega) = \frac{12}{0.294 \frac{s-1}{s+1} + 12}$$

$$\frac{T}{2} \sim \frac{1}{50} = \underline{0.02}$$



Z-transform

$$\mathcal{L}\{f(t)\} \triangleq F(s) = \int_0^{\infty} f(t) e^{-st} dt \quad \begin{cases} \text{conv} \rightarrow x \\ s \rightarrow n + i.c.'s \end{cases}$$

$$\mathcal{Z}\{f(kT)\} \triangleq F(z) = \sum_0^{\infty} f_k z^{-k} \quad \begin{cases} \text{conv} \rightarrow x \\ x_{k-1} = z^{-1} x_k \end{cases}$$

↑
 f_k



δ impulse $Z\{\delta\} = 1$.

δ unit pulse $Z\{\delta\} = \sum_0^{\infty} \delta z^{-k} = 1z^0 + 0z^{-1} + 0z^{-2} + \dots = 1$.

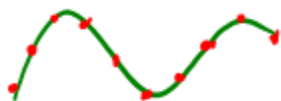
γ step $Z\{\gamma\} = \frac{1}{z}$

$Z\{\gamma\} = 1z^0 + 1z^{-1} + 1z^{-2} + \dots = \sum_0^{\infty} z^{-k}$

$$\sum_0^{\infty} \alpha^k = \frac{1}{1-\alpha} \quad + \quad |\alpha| < 1$$

$$\alpha = z^{-1} \quad \sum_0^{\infty} z^{-k} = \frac{1}{1-z^{-1}} = \left(\frac{z}{z-1} \right) = Z\{\gamma\}$$



$f(t)$ $Z\{f(t)\}$ $Z\{f(kT)\}$ 

inverse Z-transform $Z^{-1}\{z\} \rightarrow f(t)$

$Z^{-1}\{z\} \rightarrow f_k$

$f(kT)$

$1 + 0z^{-1} + 0z^{-2} \dots$

$Z^{-1}\left\{\frac{z}{z^{-1}}\right\} \rightarrow$

$$\begin{array}{r} z^{-1} \mid z \\ \hline z^{-1} \\ \hline 1 - z^{-1} \\ \hline z^{-1} \end{array}$$



Prove that \bar{z}^{-1} is a unit delay.

$$\mathcal{Z}\{f_{k-1}\} = \sum_{k=0}^{\infty} f(k-1) \bar{z}^{-k} \quad \text{let } j = k-1$$

$$= \sum_{j=0}^{\infty} f(j) \bar{z}^{-(j+1)} = \sum_{j=0}^{\infty} f(j) \bar{z}^{-j} \bar{z}^{-1}$$

$$= \bar{z}^{-1} \underbrace{\sum_{j=0}^{\infty} f(j) \bar{z}^{-j}}$$

$\mathcal{F}(z)$

$$= (\bar{z}^{-1}) \mathcal{F}(z)$$



MATLAB CODE

```
w=[0.1:1:10]';  
T=2*pi/100;  
a=1;  
n=a;  
d=[1 a];  
n1=[a*T/(1+a*T) 0];  
d1=[1 -1/(1+a*T)];  
n2 = [0 a*T/(1+a*T)];  
n3 = a*T/(2+a*T)*[1 1];  
d3 = [1 -(1-a*T/2)/(1+a*T/2)];  
bode(n,d,w); hold on  
dbode(n1,d1,T,w);  
dbode(n2,d1,T,w);  
dbode(n3,d3,T,w);  
legend('Exact','Backwards','Forwards','Tustin');
```

$$\left(\frac{a}{n+a}\right)$$

$$\frac{z^{-2}}{z^2+z+5}$$

$$[0 \ 1 \ 2], [1 \ 2 \ 5]$$



