

UNIVERSITY OF CALIFORNIA, SANTA CRUZ  
BOARD OF STUDIES IN COMPUTER ENGINEERING



CMPE-242:  
APPLIED FEEDBACK CONTROL

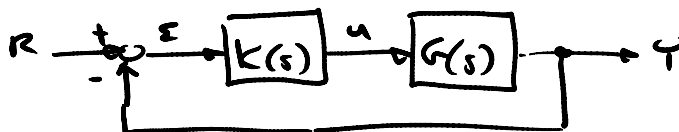
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PRACTICE MIDTERM  
MIDTERM TENTATIVELY SCHEDULED FOR 09-NOV-2010

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1. *Digital and Continuous Equivalents (25 points)*: Consider the following simple system:

$$G(s) = \frac{25}{(s + 5)^2}$$



- Design a continuous compensation system  $K(s)$  to achieve a closed loop bandwidth of 10 rad/s and a phase margin of 50 degrees.
- Convert that compensator to  $K(z)$  using a Tustin transformation and a sample time rate of 100Hz., and write down the difference equation you would implement.
- Use the exact linear phase loss from the ZOH and redesign the controller to account for it.
- Convert the plant to  $G(z)$  using a ZOH approximation, again using a sample rate of 100Hz.
- Design directly in the digital domain to achieve closed loop poles at  $s_{des} = -7 \pm 7j$
- Compare the compensator roots for all three compensators (in the continuous domain).

Problem 2 (20 points). The bode plot for the transfer function

$$G(s) = \frac{20}{s(s+1)(s+4)}$$

is shown in Figure 2. The plot shows that the system has a gain margin of 1, or 0dB, measured at  $\omega = 2 \text{ rad/s}$ ; and a phase margin of 0 at  $\omega = 2 \text{ rad/s}$ .

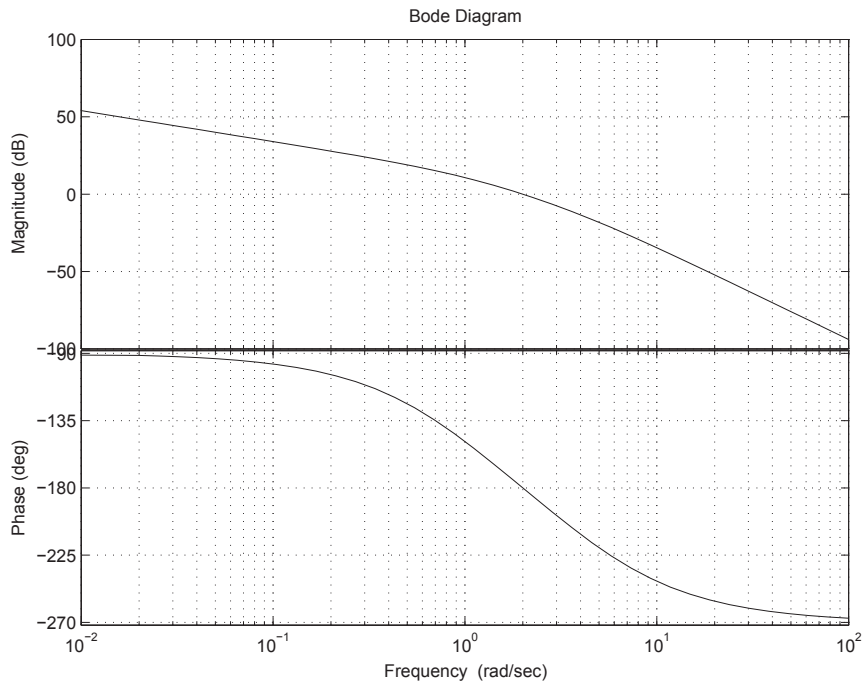


Figure 2: Bode plot for Problem 1.

- (a) (5 points). Given this Bode plot, what can you say about the locations of two of the closed loop poles of the system resulting from putting  $G(s)$  into a unity feedback loop (Figure 1) with  $K(s) = K = 1$ ?
- (b) (15 points). Design a compensator  $K(s)$  such that, with  $G(s)$  defined as above, the closed loop system shown in Figure 1 has a bandwidth of approximately 2 rad/s, steady state error to unit ramp inputs of no more than 0.1, and so that open loop transfer function  $K(s)G(s)$  has phase margin of 30.

Problem 3 (20 points).

Consider the plant  $G(s)$ , whose transfer function is given by:

$$G(s) = \frac{(s - 1)}{(s - 2)}$$

(a) (4 points). Sketch the locus of poles of the closed loop system shown in Figure 1, for a controller  $K(s)$  which is simply a positive gain  $K$ .

(b) (6 points). Sketch the locus of poles of the closed loop system shown in Figure 1, for the proper (at least as many poles as zeros), unstable controller  $K(s) = \frac{1}{s-3}$ . Hence, could this controller in series with a proportional gain  $K$  be used to stabilize the system?

(c) (4 points). Could the controller  $K(s) = \frac{1}{(s-3)(s-4)}$  in series with a proportional gain  $K$  be used to stabilize the system?

(d) (4 points). Using the same unity feedback form (Figure 1), could you stabilize  $G(s)$  with a proper, stable compensator  $K(s)$ ? Explain why or why not.

(e) (2 points). Suppose that you are given an unstable plant  $G(s)$ , which has one real zero in the right half plane. Hypothesize a condition on the possible locations of the real unstable poles of  $G(s)$  with respect to its right half plane zero, so that a proper, stable compensator  $K(s)$ , used in unity feedback (Figure 1), will stabilize the system.

Problem 4 (15 points).

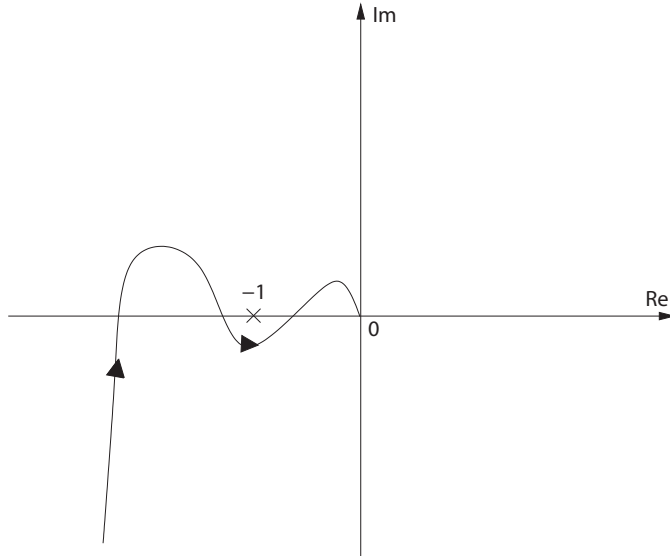


Figure 3: Nyquist plot for Problem 3.

A Nyquist plot of a unity feedback system with the feedforward transfer function  $G(s)$  is shown in Figure 3. The plot shows the curve  $G(j\omega)$ , and the arrows indicate the direction along  $G(j\omega)$  as  $\omega$  ranges from very small (positive) values to very large (positive) values. For small  $\omega$ , the curve starts in the third quadrant, though it is not specified exactly where.

(a) (8 points) Suppose that  $G(s)$  has only one pole in the closed right half  $s$ -plane (ie. the pole could be on the  $j\omega$ -axis). Complete the Nyquist plot above. Is the closed loop system (of Figure 1 with  $K = 1$ ) stable?

(b) (7 points) Now, suppose that  $G(s)$  has no pole in the closed right half  $s$ -plane, and has only one zero in the right half  $s$ -plane. Complete the Nyquist plot above. Is the closed loop system stable?